## RadTolerant Field Programmable Gate Arrays

## Features

## General Characteristics

- Tested Total Ionizing Dose (TID) Survivability Level
- No Single Event Latch-up Below a Minimum LET Threshold of $80 \mathrm{MeV}-\mathrm{cm} 2 / \mathrm{mg}$ for All RT Devices
- Packages: 84-Pin, 132-Pin, 172-Pin, 196-Pin, and 256-Pin Ceramic Quad Flat Pack
- Offered as E-Flow (Actel Space Level Flow) and Class B


## High Density and Performance

- 4,000 to 20,000 Gates
- 2,000 to 10,000 ASIC Equivalent Gates
- Up to 85 MHz On-Chip Performance
- Up to 228 User I/Os
- Up to Four Fast, Low-Skew Clock Networks


## Easy Logic Integration

- Non-Volatile, User Programmable
- Pin-Compatible Commercial Devices Available for Prototyping
- Highly Predictable Performance with $100 \%$ Automatic Place and Route
- $100 \%$ Resource Utilization with $100 \%$ Pin-Locking
- Secure Programming Technology Prevents Reverse Engineering and Design Theft
- Permanently Programmed for Instantaneous Operation on Power-Up
- Unique In-System Diagnostic and Debug Facility with Silicon Explorer
- Actel Designer Series Design Tools, Supported by Cadence, Exemplar, Mentor Graphics, Model Tech, Synopsys, Synplicity and Viewlogic Design Entry and Simulation Tools


## General Description

Actel builds the most reliable field programmable gate arrays (FPGAs) in the industry, with overall antifuse reliability ratings of less than 10 failures-in-time (FITs), corresponding to a useful life of more than 40 years. Actel FPGAs are production-proven, with more than five million devices shipped and more than one trillion antifuses manufactured. Actel devices are fully tested prior to shipment, with an out-going defect level of only 122 ppm . (Further reliability data is available in the "Actel Device Reliability Report" at http://www.actel.com/products/devices/hireldev.html.)

## Product Family Profile

| Device | RT1020 | RT1280A | RT1425A | RT1460A | RT14100A |
| :--- | ---: | ---: | ---: | ---: | ---: |
| Gates | 4,000 | 16,000 | 5,000 | 12,000 | 20,000 |
| ASIC Equivalent Gates | 2,000 | 8,000 | 2,500 | 6,000 | 10,000 |
| PLD Equivalent Gates | 5,000 | 20,000 | 6,250 | 15,000 | 25,000 |
| TTL Equivalent Package | 50 | 200 | 60 | 150 | 250 |
| 20-Pin PAL Equivalent Packages | 20 | 80 | 25 | 60 | 100 |
| Logic Modules | 547 | 1,232 | 310 | 848 | 1,377 |
| S-Modules | N/A | 624 | 160 | 432 | 697 |
| C-Modules | 547 | 608 | 150 | 416 | 680 |
| User I/Os | 69 | 140 | 100 | 168 | 228 |
| CQFP Package Pin Count | 84 | 172 | 132 | 196 | 256 |
| Performance System Speed (Maximum) | 20 MHz | 40 MHz | 60 MHz | 60 MHz | 60 MHz |
| Ordering Information |  |  |  |  |  |
| Part Number (Class B) | RT1020 | RT1280A | RT1425A | RT1460A | RT14100A |
| Part Number (E-Flow) | - CQ84B | - CQ172B | - CQ132B | -CQ196B | -CQ256B |
| Commercial Equivalent for Prototyping | A1020B | A1280A | A1425A | A1460A | A14100A |
|  | - CQ84C | - CQ172C | - CQ132C | - -CQ196C | -CQ256C |

Additionally, the programmable architecture of these devices offer high performance, design flexibility, and fast and inexpensive prototyping-all without the expense of test vectors, NRE charges, long lead times, and schedule and cost penalties for design refinements.
The RT1020 device is from the A1020B design in the ACT 1 family. This device uses a combinatorial module architecture and has 2,000 ASIC equivalent gates and 69 user I/Os. The RT1020 device is fully pin- and function-capable with the commercially-equivalent A1020B-CQ84C device for easy and inexpensive Prototyping.

The RT1280A device uses the A1280A die from the ACT 2 Family of Actel FPGAs. It utilizes a two-module architecture, consisting of combinatorial modules (C-modules) and sequential modules (S-modules) optimized for both combinatorial and sequential designs. Based on Actel's patented channeled array architecture, the RT1280A has 8,000 ASIC-equivalent gates and 140 user I/Os.
The RT1280A device is fully pin- and function-compatible with the commercially-equivalent A1280A-CQ172C device for easy and inexpensive prototyping.
The RT1425A, RT1460A and RT14100A devices use the A1425A, A1460A and A14100A die, respectively. These devices are from the ACT 3 Family of Actel FPGAs, which also utilize the two-module channeled array architecture, and offer faster performance than the RT1280A. The RT1425A has 2,500 ASIC equivalent gates and 100 user I/Os, the RT1460A has 6,000 ASIC equivalent gates and 168 user I/Os, and the RT14100A has 10,000 ASIC equivalent gates and 228 user I/Os.

These devices also have fully pin- and function-compatible commercially-equivalent devices for easy and inexpensive prototyping. The A1425A-CQ132C is used for the RT1425A, the A1460A-CQ196C is used for the RT1460A, and the A14100A-CQ256C is used for the RT14100A.

## Radiation Survivability

Total dose results are summarized in two ways. First, the maximum total dose level that is reached when the parts fail to meet a device specification but remain functional. For Actel FPGAs, the parameter that exceeds the specification first is ICC, the standby supply current. Second, the maximum total dose that is reached prior to the functional failure of the device.

The RT devices have varying total dose radiation survivability. The ability of these devices to survive radiation effects is both device and lot dependent. The customer must evaluate and determine the applicability of these devices to their specific design and environmental requirements.

Typical results for the RT1020 device has shown $\sim 100$ Krads ( Si ) for standby $\mathrm{I}_{\mathrm{CC}}$ and $>100 \mathrm{Krads}$ for functional failure. RT1280A device have shown results from 4 to 10 Krads ( Si ) for standby $I_{\mathrm{CC}}$, and 7 to 18 Krads for functional failure. ACT 3 devices typical results have shown 10 to 28 Krads for $\mathrm{I}_{\mathrm{CC}}$, and 20 to 77 Krads for functional failure.

Actel will provide Group E total dose testing on each lot that is available for sale. Actel will provide these reports on our website or you can contact your local sales representative to receive a copy. We will also provide a listing of available lots and devices. These results are only provided for reference and for customer information.

A summary of the radiation performance of Actel products ("Radiation Performance of Actel Products") can be found on the Actel Web site at
http://www.actel.com/products/devices/radhard/radperf.pdf
This summary will also show SEU and SEL testing that has been performed.

## Disclaimer

All radiation performance information is provided for information purposes only and is not guaranteed. The total dose effects are lot-dependent, and Actel does not warrant that future devices will continue to exhibit similar radiation characteristics. In addition, actual performance can vary widely due to a variety of factors, including but not limited to, characteristics of the orbit, radiation environment, proximity to satellite exterior, amount of inherent shielding from other sources within the satellite and actual bare die variations. For these reasons, Actel does not warrant any level of radiation survivability, and it is solely the responsibility of the customer to determine whether the device will meet the requirements of the specific design.

## Design Tool Support

As with all Actel FPGAs, these devices are fully supported by Actel's Designer Series development tools, which include:

- DirectTime for automated, timing-driven place and route;
- ACTgen for fast development using a wide range of macro functions; and
- ACTmap for logic synthesis.

Designer Series supports industry-leading VHDL- and Verilog-based design tools, including synthesis tools from industry leaders such as Exemplar Logic, Synplicity, and Synopsys. ${ }^{1}$

[^0]In addition, these devices are supported by Actel's new Silicon Explorer diagnostic and debugging tool kit. Silicon Explorer dramatically reduces verification time from several hours per cycle to a few seconds by enabling real-time, in-circuit debugging. Silicon Explorer includes:

- Probe Pilot, a high-speed signal acquisition and control tool that samples data at 100 MHz (asynchronous) or 66 MHz (synchronous). Probe Pilot features 18 probing channels and connects to the user's PC via a standard serial port connection.
- Diagnostic software, which turns the PC into a fully-featured, 100 MHz logic analyzer for easy graphical analysis of waveforms.
Silicon Explorer probes 100 percent of the device circuitry using Probe Pilot's powerful, 18-channel signal acquisition capability. Individual bugs are then isolated and passed to the user interface, providing the user with complete waveform data.


## RadTolerant Device Ordering Information



## Product Plan

|  | Application |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | C | M | B | E |
| ACT 2 |  |  |  |  |
| RT1020 Device |  |  |  |  |
| 84-Pin Ceramic Quad Flat Pack (CQFP) | - | - | $\checkmark$ | $\checkmark$ |
| A1020B Device (Prototyping Use) |  |  |  |  |
| 84-Pin Ceramic Quad Flat Pack (CQFP) | $\checkmark$ | $\checkmark$ | $\checkmark$ | - |
| RT1280A Device |  |  |  |  |
| 172-Pin Ceramic Quad Flat Pack (CQFP) | - | - | $\checkmark$ | $\checkmark$ |
| A1280A Device (Prototyping Use) |  |  |  |  |
| 172-Pin Ceramic Quad Flat Pack (CQFP) | $\checkmark$ | $\checkmark$ | $\checkmark$ | - |
| ACT 3 |  |  |  |  |
| RT1425A Device |  |  |  |  |
| 132-Pin Ceramic Quad Flat Pack (CQFP) | - | - | $\checkmark$ | $\checkmark$ |
| A1425A Device (Prototyping Use) |  |  |  |  |
| 132-Pin Ceramic Quad Flat Pack (CQFP) | $\checkmark$ | $\checkmark$ | $\checkmark$ | - |
| RT1460A Device |  |  |  |  |
| 196-Pin Ceramic Quad Flat Pack (CQFP) | - | - | $\checkmark$ | $\checkmark$ |
| A1460A Device (Prototyping Use) |  |  |  |  |
| 196-Pin Ceramic Quad Flat Pack (CQFP) | $\checkmark$ | $\checkmark$ | $\checkmark$ | - |
| RT14100A Device |  |  |  |  |
| 256-Pin Ceramic Quad Flat Pack (CQFP) | - | - | $\checkmark$ | $\checkmark$ |
| A14100A Device (Prototyping Use) |  |  |  |  |
| 172-Pin Ceramic Quad Flat Pack (CQFP) | $\checkmark$ | $\checkmark$ | $\checkmark$ | - |
| Applications: $\mathrm{C}=$ Commercial <br>  $\mathrm{M}=$ Military <br>  $\mathrm{B}=$ MIL-STD-883 Class B <br>  $\mathrm{E}=$ Extended Flow (Space Level) |  |  |  |  |

## Device Resources

|  |  |  | User I/Os |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FPGA Device Type |  |  |  |  |  |  |  |

## Architectural Overview

The Actel architecture is composed of fine-grained logic modules which produce fast, efficient logic designs. All devices are composed of logic modules, routing resources, clock networks, and I/O modules which are the building blocks for fast logic designs.

## Logic Modules

These devices contain two types of logic modules: combinatorial ( C -modules) and sequential ( S -modules). RT1020 and A1020B devices contain only C-modules.
The C-module, shown in Figure 1, implements the following function:

$$
Y=!S 1 *!50 * D 00+!S 1 * 50 * D 01+S 1 *!50 * D 10+S 1 * 50 * D 11
$$

where:

$$
\begin{aligned}
& \mathrm{S} 0=\mathrm{A} 0 * \mathrm{~B} 0 \\
& \mathrm{~S} 1=\mathrm{A} 1+\mathrm{B} 1
\end{aligned}
$$

The S -module shown in Figure 2 is designed to implement high-speed sequential functions within a single logic module. The S-module implements the same combinatorial logic


Figure 1 - C-Module Implementation
function as the C -module while adding a sequential element. The sequential element can be configured as either a D-type flip-flop or a transparent latch. To increase flexibility, the $S$-module register can be by-passed so it implements purely combinatorial logic.


Up to 7-Input Function Plus D-Type Flip-Flop with Clear


Up to 4-Input Function Plus Latch with Clear


Up to 7-Input Function Plus Latch


Up to 8-Input Function (Same as C-Module)

Figure 2 •S-Module Implementation

Flip-flops can also be created using two C -modules. The single event upset (SEU) characteristics differ between an S-module flip-flop and a flip-flop created using two C-modules. See the Radiation Specifications in this Data Sheet for details and the Actel Application Note, "Design Techniques for RadHard Field Programmable Gate Arrays" found at http://www.actel.com/products/radhard.html.

## The ACT 1 Logic Module

The ACT 1 logic module is an 8-input, one-output logic circuit chosen for the wide range of functions it implements and for its efficient use of interconnect routing resources (Figure 3).

The logic module can implement the four basic logic functions (NAND, AND, OR, and NOR) in gates of two, three, or four inputs. Each function may have many versions, with different combinations of active-low inputs. The logic module can also implement a variety of D-latches, exclusivity functions, AND-ORs, and OR-ANDs. No dedicated hardwired latches or flip-flops are required in the array, since latches and flip-flops may be constructed from logic modules wherever needed in the application.


Figure 3 - ACT 1 Logic Module

## I/O Modules

I/O modules provide the interface between the device pins and the logic array; Figure 4 is a block diagram of the I/O module. A variety of user functions, determined by a library macro selection, can be implemented in the module (refer to the Macro Library Guide for more information). I/O modules contain a tri-state buffer, and input and output latches which can be configured for input, output, or bi-directional pins (Figure 4).


Figure 4 • I/O Module
The RadHard devices contain flexible I/O structures in that each output pin has a dedicated output enable control. The I/O module can be used to latch input and/or output data, providing a fast set-up time. In addition, the Actel Designer Series software tools can build a D flip-flop, using a C-module, to register input and/or output signals.
Actel's Designer Series development tools provide a design library of I/O macros. The I/O macro library provides macro functions that can implement all I/O configurations supported by the RadHard FPGAs.

## Routing Structure

The RadHard device architecture uses vertical and horizontal routing tracks to interconnect the various logic and I/O modules. These routing tracks are metal interconnects that may either be of continuous length or broken into pieces called segments. Varying segment lengths allows over $90 \%$ of the circuit interconnects to be made with only two antifuse connections. Segments can be joined together at the ends, using antifuses to increase their lengths up to the full length of the track. All interconnects can be accomplished with a maximum of four antifuses.

## Horizontal R outing

Horizontal channels are located between the rows of modules, and are composed of several routing tracks. The horizontal routing tracks within the channel are divided into one or more segments. The minimum horizontal segment length is the width of a module-pair, and the maximum horizontal segment length is the full length of the channel. Any segment that spans more than one-third the row length is considered a
long horizontal segment. A typical channel is shown in Figure 5. Non-dedicated horizontal routing tracks are used to route signal nets. Dedicated routing tracks are used for the global clock networks, and for power and ground tie-off tracks.

## Vertical Routing

Another set of routing tracks run vertically through the module. Vertical tracks are of three types: input, output, and long. Vertical tracks are also divided into one or more segments. Each segment in an input track is dedicated to the input of a particular module. Each segment in an output track is dedicated to the output of a particular module. Long segments are uncommitted and can be assigned during routing. Each output segment spans four channels (two above and two below), except near the top and bottom of the array where edge effects occur. LVTs contain either one or two segments. An example of vertical routing tracks and segments is shown in Figure 5.


Figure 5 • Routing Structure

## Antifuse Structures

An antifuse is a "normally open" structure as opposed to the normally closed fuse structure used in PROMs or PALs. The use of antifuses to implement a programmable logic device results in highly testable structures, as well as efficient programming algorithms. The structure is highly testable because there are no pre-existing connections; therefore, temporary connections can be made using pass transistors. These temporary connections can isolate individual antifuses to be programmed, as well as isolate individual circuit structures to be tested. This can be done both before and after programming. For example, all metal tracks can be tested for continuity and shorts between adjacent tracks, and the functionality of all logic modules can be verified.

## Pin Description

## CLK Clock (Input)

RT1020 and A1020B only. TTL clock input for global clock distribution networks. The clock input is buffered prior to clocking the logic modules. This pin can also be used as an I/O.

## CLKA Clock A (Input)

Not applicable for RT1020 and A1020B. TTL clock input for global clock distribution networks. The clock input is buffered prior to clocking the logic modules. This pin can also be used as an I/O.

## CLKB Clock B (Input)

Not applicable for RT1020 and A1020B. TTL clock input for global clock distribution networks. The clock input is buffered prior to clocking the logic modules. This pin can also be used as an $\mathrm{I} / 0$.

## DCLK Diagnostic Clock (Input)

TTL clock input for diagnostic probe and device programming. DCLK is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

## GND <br> Ground

LOW supply voltage.

## HCLK Dedicated (Hard-Wired) Array Clock (Input)

Not applicable for RT1020, A1020B, RT1280A and A1280A. TTL clock input for sequential modules. This input is directly wired to each S-module, offering clock speeds independent of the number of S -modules being driven. This pin can also be used as an I/O.

## I/O Input/Output (Input, Output)

I/O pin functions as an input, output, tri-state, or bi-directional buffer. Input and output levels are compatible with standard TTL and CMOS specifications. Unused I/O pins are automatically driven LOW.

## IOCLK <br> Dedicated (Hard-Wired) I/O Clock (Input)

Not applicable for RT1020, A1020B, RT1280A and A1280A. TTL clock input for I/O modules. This input is directly wired to each I/O module, offering clock speeds independent of the number of I/O modules being driven. This pin can also be used as an I/O.

## IOPCL Dedicated (Hard-Wired) I/O Preset/Clear (Input)

Not applicable for RT1020, A1020B, RT1280A and A1280A. TTL input for I/O preset or clear. This global input is directly wired to the preset and clear inputs of all I/O registers. This pin functions as an I/O when no I/O preset or clear macros are used.

## MODE Mode (Input)

The MODE pin controls the use of diagnostic pins (DCLK, PRA, PRB, SDI). When the MODE pin is HIGH, the special functions are active. When the MODE pin is LOW, the pins function as I/Os. To provide ActionProbe capability, the MODE pin should be terminated to GND through a 10K resistor so that the MODE pin can be pulled HIGH when required.

## NC

## No Connection

This pin is not connected to circuitry within the device.

## PRA, I/O Probe A (Output)

The Probe A pin is used to output data from any user-defined design node within the device. This independent diagnostic pin can be used in conjunction with the Probe B pin to allow real-time diagnostic output of any signal path within the device. The Probe A pin can be used as a user-defined I/O when debugging has been completed. PRA is accessible when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

## PRB, I/O Probe B (Output)

The Probe B pin is used to output data from any user-defined design node within the device. This independent diagnostic pin can be used in conjunction with the Probe A pin to allow real-time diagnostic output of any signal path within the device. The Probe B pin can be used as a user-defined I/O when debugging has been completed. PRB is accessible when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

## SDI Serial Data Input (Input)

Serial data input for diagnostic probe and device programming. SDI is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

```
Vcc 5V Supply Voltage
HIGH supply voltage.
```


## Actel MIL-STD-883 Product Flow

| Step | Screen | 883 Method | 883-Class B Requirement |
| :---: | :---: | :---: | :---: |
| 1.0 | Internal Visual | 2010, Test Condition B | 100\% |
| 2.0 | Temperature Cycling | 1010, Test Condition C | 100\% |
| 3.0 | Constant Acceleration | 2001, Test Condition E (Min), Y1, Orientation Only | 100\% |
| 4.0 | Seal <br> a. Fine <br> b. Gross | 1014 | $\begin{aligned} & 100 \% \\ & 100 \% \end{aligned}$ |
| 5.0 | Visual Inspection | 2009 | 100\% |
| 6.0 | Pre-Burn-In Electrical Parameters | In accordance with applicable Actel device specification | 100\% |
| 7.0 | Burn-in Test | 1015 Condition D 160 hours @ $125^{\circ} \mathrm{C}$ Min. | 100\% |
| 8.0 | Interim (Post-Burn-In) Electrical Parameters | In accordance with applicable Actel device specification | 100\% |
| 9.0 | Percent Defective Allowable | 5\% | All Lots |
| 10.0 | Final Electrical Test <br> a. Static Tests <br> (1) $25^{\circ} \mathrm{C}$ <br> (Subgroup 1, Table I, 5005) <br> (2) $-55^{\circ} \mathrm{C}$ and $+125^{\circ} \mathrm{C}$ <br> (Subgroups 2, 3, Table I, 5005) <br> b. Dynamic and Functional Tests <br> (1) $25^{\circ} \mathrm{C}$ <br> (Subgroup 7, Table I, 5005) <br> (2) $-55^{\circ} \mathrm{C}$ and $+125^{\circ} \mathrm{C}$ <br> (Subgroups 8A and 8B, Table I, 5005) <br> c. Switching Tests at $25^{\circ} \mathrm{C}$ <br> (Subgroup 9, Table I, 5005) | In accordance with applicable Actel device specification | 100\% <br> 100\% <br> 100\% |
| 11.0 | Qualification or Quality Confirmation Inspection Test Sample Selection (Group A and Group B) | 5005 | All Lots |
| 12.0 | External Visual | 2009 | 100\% |

## Actel Extended Flow ${ }^{1}$

| Step | Screen | Method | Requirement |
| :---: | :---: | :---: | :---: |
| 1. | Wafer Lot Acceptance ${ }^{2}$ | 5007 with Step Coverage Waiver | All Lots |
| 2. | Destructive In-Line Bond Pull ${ }^{3}$ | 2011, Condition D | Sample |
| 3. | Internal Visual | 2010, Condition A | 100\% |
| 4. | Serialization |  | 100\% |
| 5. | Temperature Cycling | 1010, Condition C | 100\% |
| 6. | Constant Acceleration | 2001, Condition E (Min), Y ${ }_{1}$ Orientation Only | 100\% |
| 7. | Visual Inspection | 2009 | 100\% |
| 8. | Particle Impact Noise Detection | 2020, Condition A | 100\% |
| 9. | Radiographic | 2012 | 100\% |
| 10. | Pre-Burn-In Test | In accordance with applicable Actel device specification | 100\% |
| 11. | Burn-in Test | 1015, Condition D, 240 hours @ $125^{\circ} \mathrm{C}$ minimum | 100\% |
| 12. | Interim (Post-Burn-In) Electrical Parameters | In accordance with applicable Actel device specification | 100\% |
| 13. | Reverse Bias Burn-In | 1015, Condition C, 72 hours @ $150^{\circ} \mathrm{C}$ minimum | 100\% |
| 14. | Interim (Post-Burn-In) Electrical Parameters | In accordance with applicable Actel device specification | 100\% |
| 15. | Percent Defective Allowable (PDA) Calculation | 5\%, 3\% Functional Parameters @ $25^{\circ} \mathrm{C}$ | All Lots |
|  | Final Electrical Test <br> a. Static Tests <br> (1) $25^{\circ} \mathrm{C}$ (Subgroup 1, Table1) <br> (2) $-55^{\circ} \mathrm{C}$ and $+125^{\circ} \mathrm{C}$ (Subgroups 2, 3, Table 1) <br> b. Dynamic and Functional Tests <br> (1) $25^{\circ} \mathrm{C}$ (Subgroup 7, Table 15) <br> (2) $-55^{\circ} \mathrm{C}$ and $+125^{\circ} \mathrm{C}$ (Subgroups 5 and 6, 8a and b, Table 1) <br> c. Switching Tests at $25^{\circ} \mathrm{C}$ (Subgroup 9, Table 1) | In accordance with Actel applicable device specification 5005 5005 5005 5005 5005 | $\begin{aligned} & 100 \% \\ & 100 \% \\ & 100 \% \\ & 100 \% \end{aligned}$ |
|  | Seal <br> a. Fine <br> b. Gross | 1014 | 100\% |
| 18. | Qualification or Quality Conformance Inspection Test Sample Selection | 5005 | Group A \& Group B |
| 19 | External Visual | 2009 | 100\% |

## Notes:

1. Actel offers the extended flow for customers that require additional screening beyond the requirements of MIL-STD-883, Class B. Actel is compliant to the requirements of MIL-STD-883, Paragraph 1.2.1, and MIL-I-38535, Appendix A. Actel is offering this extended flow incorporating the majority of the screening procedures as outlined in Method 5004 of MIL-STD-883 Class S. The excepti ons to Method 5004 areshown in notes 2 to 3 below.
2. Wafer lot acceptance is performed to Method 5007; however, the step coverage requirement as specifi ed in Method 2018 must be waived.
3. Method 5004 requires a 100 percent, non-destructive bond pull to Method 2023. Actel substitutes a destructive bond pull to Method 2011, Condition D on a sample basis only.

Absolute Maximum Ratings ${ }^{1}$
Free Air Temperature Range

| Symbol | Parameter | Limits | Units |
| :--- | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{CC}}$ | DC Supply Voltage $^{2}$ | -0.5 to +7.0 | V |
| $\mathrm{~V}_{\mathrm{I}}$ | Input Voltage | -0.5 to $\mathrm{V}_{\mathrm{CC}}+0.5$ | V |
| $\mathrm{~V}_{\mathrm{O}}$ | Output Voltage $^{\text {I/O Source Sink }}$ | -0.5 to $\mathrm{V}_{\mathrm{CC}}+0.5$ | V |
| $\mathrm{I}_{\mathrm{IO}}$ | Current |  |  |
| $\mathrm{T}_{\text {STG }}$ | Storage Temperature | -65 to +150 | mA |

## Notes:

1. Stresses beyond thoselisted under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect devicereliability. Device should not beoperated outsidethe recommended operating conditions.
2. Device inputs are normally high impedance and draw extremely low current. However, when input voltage is greater than $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ or less than GND -0.5 V , the internal protection diodewill be forward-biased and can draw excessivecurrent.

## Recommended Operating Conditions

| Parameter | Commercial | Military | Units |
| :--- | :--- | :--- | :--- |
| Temperature <br> Range $^{1}$ | 0 to +70 | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |
| Power Supply <br> Tolerance | $\pm 5$ | $\pm 10$ | $\% \mathrm{~V}_{\mathrm{CC}}$ |

## Note:

1. Ambient temperature $\left(T_{A}\right)$ is used for commercial and industrial; case temperature $\left(T_{C}\right)$ is used for military.

## Electrical Specifications



## Notes:

1. Actel devi ces can driveand receiveei ther CMOS or TTL signal levels. No assi gnment of I/Os as TTL or CMOS is required.
2. Tested one output at a time, $\mathrm{V}_{\mathrm{CC}}=\mathrm{min}$.
3. Not tested; for information only.
4. $V_{\text {OUT }}=0 V, f=1 \mathrm{MHz}$

## Package Thermal Characteristics

The device junction to case thermal characteristic is $\theta_{j \mathrm{j}}$, and the junction to ambient air characteristic is $\theta_{\mathrm{id}}$. The thermal characteristics for $\theta_{\mathrm{ja}}$ are shown with two different air flow rates.

Maximum junction temperature is $150^{\circ} \mathrm{C}$.
A sample calculation of the absolute maximum power dissipation allowed for a CQFP 172-pin package at military temperature is as follows:

$$
\frac{\text { Max. junction temp. }\left({ }^{\circ} \mathrm{C}\right)-\text { Max. military temp. }}{\theta_{\mathrm{jA}}\left({ }^{\circ} \mathrm{C} / \mathrm{W}\right)}=\frac{150^{\circ} \mathrm{C}-125^{\circ} \mathrm{C}}{25^{\circ} \mathrm{C} / \mathrm{W}}=1.0 \mathrm{~W}
$$

| Package Type | Pin Count | $\theta_{\text {jc }}$ | $\theta_{\text {ja }}$ <br> Still | $\theta_{\mathbf{j a r}}$ <br> $\mathbf{3 0 0} \mathbf{f t / m i n}$ | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Ceramic Quad Flat Pack | 84 | 5 | 40 | 30 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | 132 | 13 | 55 | 30 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | 172 | 13 | 25 | 15 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | 196 | 13 | 36 | 24 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | 256 | 13 | 30 | 18 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## General Power Equation

$$
P=\left[I_{C C} \text { standby }+I_{C C} \text { active }\right] * V_{C C}+I_{O L} * V_{O L} * N+
$$

$$
\mathrm{I}_{\mathrm{OH}} *\left(\mathrm{~V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{OH}}\right) * \mathrm{M}
$$

## Where:

${ }^{\text {CC }}$ standby is the current flowing when no inputs or outputs are changing.
${ }^{\text {CC }}$ active is the current flowing due to CMOS switching.
$I_{\text {OL }}, I_{\text {OH }}$ are TTL sink/source currents.
$\mathrm{V}_{\mathrm{OL}}, \mathrm{V}_{\mathrm{OH}}$ are TTL level output voltages.
$N$ equals the number of outputs driving TTL loads to $\mathrm{V}_{\mathrm{OL}}$.
$M$ equals the number of outputs driving TTL loads to $\mathrm{V}_{\mathrm{OH}}$.
An accurate determination of $N$ and $M$ is problematical because their values depend on the family type, on design details, and on the system I/O. The power can be divided into two components: static and active.

## Static Power Component

Actel FPGAs have small static power components that result in power dissipation lower than that of PALs or PLDs. By integrating multiple PALs or PLDs into one FPGA, an even greater reduction in board-level power dissipation can be achieved.
The power due to standby current is typically a small component of the overall power. Standby power is calculated below for commercial, worst-case conditions.

| $I_{\text {CC }}$ | $V_{\text {CC }}$ | Power |
| :---: | :---: | :---: |
| 2 mA | 5.25 V | 10.5 mW |

The static power dissipated by TTL loads depends on the number of outputs driving HIGH or LOW and on the DC load current. Again, this value is typically small. For instance, a 32-bit bus sinking 4 mA at 0.33 V will generate 42 mW with all outputs driving LOW, and 140 mW with all outputs driving HIGH.

## Active Power Component

Power dissipation in CMOS devices is usually dominated by the active (dynamic) power dissipation. This component is frequency-dependent, a function of the logic and the external I/O. Active power dissipation results from charging internal chip capacitances of the interconnect, unprogrammed antifuses, module inputs, and module outputs, plus external capacitance due to PC board traces and load device inputs. An additional component of the active power dissipation is the totem pole current in CMOS transistor pairs. The net effect can be associated with an equivalent capacitance that can be combined with frequency and voltage to represent active power dissipation.

## Equivalent Capacitance

The power dissipated by a CMOS circuit can be expressed by Equation 1

$$
\begin{equation*}
\text { Power (uW) }=C_{E Q} * V_{C C}{ }^{2} * F \tag{1}
\end{equation*}
$$

where:
$\mathrm{C}_{\mathrm{EQ}}$ is the equivalent capacitance expressed in pF .
$V_{C C}$ is the power supply in volts (V).
F is the switching frequency in MHz .

Equivalent capacitance is calculated by measuring $I_{C C}$ active at a specified frequency and voltage for each circuit component of interest. Measurements are made over a range of frequencies at a fixed value of $\mathrm{V}_{\mathrm{CC}}$. Equivalent capacitance is frequency-independent, so the results can be used over a wide range of operating conditions. Equivalent capacitance values are shown below.

## CEQ Values for Actel FPGAs

$\left.\begin{array}{lrrr} & & & \begin{array}{r}\text { RT1425A } \\ \text { A1425A }\end{array} \\ & & & \begin{array}{r}\text { RT1460A } \\ \text { A1460A }\end{array} \\ & & & \\ & \text { RT1020 } & \text { RT1280A } & \text { RT14100A }\end{array}\right]$

To calculate the active power dissipated from the complete design, the switching frequency of each part of the logic must be known. Equation 2 shows a piece-wise linear summation over all components. Since the RT1280A and A1280A have two routed array clocks, the dedicated_Clk and IO_Clk terms do not apply. For all other devices all terms will apply.

> Power $=V_{C C}{ }^{2} *\left[\left(m * C_{E Q M} * f_{m}\right)_{\text {modules }}+\left(n * C_{E Q I} * f_{n}\right)_{\text {inputs }}+\right.$ $\left(p *\left(C_{E Q O}+C_{L}\right) * f_{p}\right)_{\text {outputs }}+0.5 *\left(q_{1} * C_{E Q C R} * f_{q 1}\right)_{\text {routed_Clk1 }}$ $+\left(r_{1} * f_{q 1}\right)_{\text {routed_Clk1 }}+0.5 *\left(q_{2} * C_{E Q C R} * f_{q 2}\right)_{\text {routed_Clk2 }}{ }^{-}$ $\left(\mathrm{r}_{2} * \mathrm{f}_{\mathrm{q} 2}\right)_{\text {routed_Clk2 }}+0.5 *\left(\mathrm{~s}_{1} * \mathrm{C}_{\mathrm{EQCD}} * \mathrm{f}_{\mathrm{s} 1}\right)_{\text {dedicated_CIk }}+$ $\left.\left(\mathrm{s}_{2} * \mathrm{C}_{\mathrm{EQCl}} * \mathrm{f}_{\mathrm{S} 2}^{-}\right)_{\mathrm{IO}_{-} \mathrm{Clk}}\right]$
> (2)
where:

| m | $=$ Number of logic modules switching at $\mathrm{f}_{\mathrm{m}}$ |
| :---: | :---: |
| n | $=$ Number of input buffers switching at $f_{n}$ |
| p | $=$ Number of output buffers switching at $f_{p}$ |
| $q_{1}$ | $=$ Number of clock loads on the first routed array clock |
| $q_{2}$ | $=$ Number of clock loads on the second routed array clock |
| $\mathrm{r}_{1}$ | $=$ Fixed capacitance due to first routed array clock |
| $r_{2}$ | $=$ Fixed capacitance due to second routed array clock |
| $\mathrm{S}_{1}$ | $=$ Fixed number of clock loads on the dedicated array clock (Not applicable for RT1280A and A1280A.) |
| $S_{2}$ | $=$ Fixed number of clock loads on the dedicated I/O clock (Not applicable for RT1280A and A1280A.) |
| $\mathrm{C}_{\mathrm{EQM}}$ | $=$ Equivalent capacitance of logic modules in pF |
| $\mathrm{C}_{\text {EQI }}$ | $=$ Equivalent capacitance of input buffers in pF |
| $\mathrm{C}_{\text {EQO }}$ | $=$ Equival ent capacitance of output buffers in pF |
| $\mathrm{C}_{\text {EQCR }}$ | ```= Equivalent capacitance of routed array clock in pF``` |
| $C_{\text {EQCD }}$ | ```= Equival ent capacitance of dedicated array clock in pF``` |
| $\mathrm{C}_{\mathrm{EQCl}}$ | $\begin{aligned} & =\text { Equivalent capacitance of dedicated I/O clock } \\ & \text { in pF } \end{aligned}$ |
| $C_{L}$ | $=$ Output lead capacitance in pF |
| $f_{m}$ | $=$ Average logic module switching rate in MHz |
| $\mathrm{f}_{\mathrm{n}}$ | $=$ Average input buffer switching rate in MHz |
| $\mathrm{f}_{\mathrm{p}}$ | $=$ Average output buffer switching rate in MHz |
| $\mathrm{f}_{\mathrm{q} 1}$ | $=$ Average first routed array clock rate in MHz |
| $\mathrm{f}_{\mathrm{q} 2}$ | $=$ Average second routed array clock rate in MHz |
| $\mathrm{f}_{\mathrm{s} 1}$ | $=$ Average dedicated array clock rate in MHz <br> (Not applicable for RT1280A and A1280A.) |
| $\mathrm{f}_{\mathrm{s} 2}$ | = Average dedicated I/O clock rate in MHz <br> (Not applicable for RT1280A and A1280A.) |

## Fixed Capacitance Values for Actel FPGAs (pF)

| Device Type | $\mathbf{r}_{1}$ <br> routed_Clk1 | $\mathbf{r}_{2}$ <br> routed_Clk2 |
| :--- | :---: | :---: |
| RT1020, A1020B | 69 | $\mathrm{n} / \mathrm{a}$ |
| RT1280A, A1280A | 168 | 168 |
| RT1425A, A1425A | 75 | 75 |
| RT1460A, A1460A | 165 | 165 |
| RT14100A, A14100A | 195 | 195 |

Fixed Clock Loads ( $\mathbf{s}_{1} / \mathrm{s}_{\mathbf{2}}-\mathrm{ACT} 3$ Only)

|  | $\boldsymbol{s}_{1}$ <br> Clock Loads on <br> Dedicated | $\boldsymbol{s}_{2}$ <br> Clock Loads on <br> Dedicated <br> Array Clock |
| :--- | :---: | :---: |
| I/O Clock |  |  |

## Determining Average Switching Frequency

To determine the switching frequency for a design, you must have a detailed understanding of the data input values to the circuit. The guidelines in the table below are meant to represent worst-case scenarios; they can be generally used to predict the upper limits of power dissipation.

## RT1020, A1020B, RT1280A, A1280A

| Logic Modules (m) $=8$ | 80\%of Combinatoria Modules |
| :---: | :---: |
| InputSwitching ( n ) $=$ \# | = \#nputs/4 |
| Outputs Switching (p) = \# | = Outputs/4 |
| First Routed Array Clock Loads ( $q_{1}$ ) = 4 | $40 \%$ of Sequential Modules |
| $\begin{aligned} & \text { Second Routed Array Clock Loads }=40 \\ & \left(\mathrm{q}_{2}\right) \end{aligned}$ | = $40 \%$ of Sequential Modules |
| Load Capacitance ( $C_{L}$ ) $=35$ | $=35 \mathrm{pF}$ |
| $\begin{aligned} & \text { Average Logic Module Switching }=F \\ & \text { Rate }\left(f_{m}\right) \end{aligned}$ | $=F / 10$ |
| Average Input Switching Rate $\left(f_{n}\right)=F$ | F/5 |
| Average Output Switching Rate $\left(f_{p}\right)=F$ | F/10 |
| $\begin{aligned} & \text { Average First Routed Array Clock }=\mathrm{F} \\ & \text { Rate }\left(\mathrm{f}_{\mathrm{q} 1}\right) \end{aligned}$ | $=F$ |
| $\begin{aligned} & \text { Average Second Routed Array Clock }=\mathrm{F} / \\ & \text { Rate }\left(\mathrm{f}_{\mathrm{q} 2}\right) \end{aligned}$ | $=F / 2$ |
| Average Dedicated Array Clock Rate $=$ $\left(f_{s 1}\right)$ | $=\mathrm{n} / \mathrm{a}$ |
| $\begin{aligned} & \text { Average Dedicated I/O Clock Rate }=\mathrm{n} / \\ & \left(\mathrm{f}_{\mathrm{s} 2}\right) \end{aligned}$ | $=\mathrm{n} / \mathrm{a}$ |

RT1425A, A1425A, RT1460A, A1460A, RT14100A, A14100A

| Logic Modules $(\mathrm{m})$ | $=$$80 \%$ of Combinatorial <br>  <br> Modules |
| :--- | :--- |
| Input Switching $(\mathrm{n})$ | $=$ \#Inputs/4 |
| Outputs Switching $(\mathrm{p})$ | $=$ \#Outputs/4 |
| First Routed Array Clock Loads $\left(\mathrm{q}_{1}\right)=$ | $40 \%$ of Sequential |
|  | Modules |

## RT1020, A1020B Timing Module



## RT1280A, Al280A Timing Model*


*Values shown for RT1280A-1 at worst-case military conditions.
$\dagger$ Input module predicted routing delay

RT1425A, Al425A, RT1460A, A1460A, RT14100A, Al4100A Timing Model*

*Values shown for RT14100A-1 at worst-case military conditions.

## Parameter Measurement

## Output Buffer Delays



AC Test Load

Load 1
(Used to measure propagation delay)


Load 2
(Used to measure rising/falling edges)


## Sequential Timing Characteristics

Flip-Flops and Latches (RT1280A, A1280A)

(Positive Edge-Triggered)


## Note:

1. D represents all data functions involving $A, B$, and $S$ for multiplexed flip-flops.

## Sequential Timing Characteristics (continued)

Flip-Flops and Latches (RT1425A, A1425A, RT1460A, A1460A, RT14100A, A14100A)


Note:

1. D represents all data functions involving $A, B$, and $S$ for multiplexed flip-flops.

## Sequential Timing Characteristics (continued)

Input Buffer Latches (RT1280A, A1280A)


Output Buffer Latches (RT1280A, A1280A)


## RT1020, A1020B Timing Characteristics

(Worst-Case Military Conditions, $\mathrm{V}_{\mathrm{Cc}}=4.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{J}}=125^{\circ} \mathrm{C}$ )

| Logic Module Propagation Delays |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Units |
| $\mathrm{t}_{\text {PD1 }}$ | Single Module |  | 3.6 | ns |
| $\mathrm{t}_{\text {PD2 }}$ | Dual Module Macros |  | 8.4 | ns |
| $\mathrm{t}_{\mathrm{CO}}$ | Sequential Clock to Q |  | 3.6 | ns |
| $\mathrm{t}_{\mathrm{GO}}$ | Latch G to Q |  | 3.6 | ns |
| $\mathrm{t}_{\mathrm{RS}}$ | Flip-Flop (Latch) Reset to Q |  | 3.6 | ns |
| Predicted Routing Delays ${ }^{1}$ |  |  |  |  |
| $\mathrm{t}_{\text {RD1 }}$ | FO=1 Routing Delay |  | 1.1 | ns |
| $\mathrm{t}_{\mathrm{RD} 2}$ | FO=2 Routing Delay |  | 1.8 | ns |
| $\mathrm{t}_{\text {RD3 }}$ | FO=3 Routing Delay |  | 2.6 | ns |
| $\mathrm{t}_{\mathrm{RD} 4}$ | FO=4 Routing Delay |  | 3.9 | ns |
| $\mathrm{t}_{\text {RD8 }}$ | FO=8 Routing Delay |  | 8.1 | ns |
| Sequential Timing Characteristics ${ }^{\mathbf{2}}$ |  |  |  |  |
| $\mathrm{t}_{\text {SUD }}$ | Flip-Flop (Latch) Data Input Set-Up | 6.9 |  | ns |
| $\mathrm{t}_{\mathrm{HD}}{ }^{3}$ | Flip-Flop (Latch) Data Input Hold | 0.0 |  | ns |
| tsuena | Flip-Flop (Latch) Enable Set-Up | 6.9 |  | ns |
| thena | Flip-Flop (Latch) Enable Hold | 0.0 |  | ns |
| $t_{\text {WCLKA }}$ | Flip-Flop (Latch) Clock Active Pulse Width | 8.4 |  | ns |
| $\mathrm{t}_{\text {WASYN }}$ | Flip-Flop (Latch) Asynchronous Pulse Width | 8.4 |  | ns |
| $\mathrm{t}_{\mathrm{A}}$ | Flip-Flop Clock Input Period | 17.5 |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Flip-Flop (Latch) Clock Frequency ( $\mathrm{FO}=128$ ) |  | 55 | MHz |

## Notes:

1. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on thedevice prior to shipment.
2. Set-up times assume fanout of 3 . Further testing information can be obtained from the DirectTime Analyzer utility.
3. The hold time for the DFME1A macro may be greater than 0 ns . Use the Designer Series 3.0 (or later) Timer to check the hold time for this macro.

## RT1020, A1020B Timing Characteristics (continued) (Worst-Case Military Conditions)

| Input Module Propagation Delays |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description |  | Min. | Max. | Units |
| $\mathrm{t}_{\text {INYH }}$ | Pad to Y High |  |  | 3.9 | ns |
| tinyt | Pad to Y Low |  |  | 3.9 | ns |
| Input Module Predicted Routing Delays ${ }^{1}$ |  |  |  |  |  |
| tIRD1 | FO=1 Routing Delay |  |  | 1.1 | ns |
| $\mathrm{t}_{\text {IRD2 }}$ | FO=2 Routing Delay |  |  | 1.8 | ns |
| tIRD3 | FO=3 Routing Delay |  |  | 2.6 | ns |
| tiRD4 | $\mathrm{FO}=4$ Routing Delay |  |  | 3.9 | ns |
| $\mathrm{t}_{\text {IRD8 }}$ | FO=8 Routing Delay |  |  | 8.1 | ns |
| Global Clock Network |  |  |  |  |  |
| $\mathrm{t}_{\text {CKH }}$ | Input Low to High | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ |  | $\begin{aligned} & 6.0 \\ & 6.9 \end{aligned}$ | ns |
| $\mathrm{t}_{\mathrm{CKL}}$ | Input High to Low | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ |  | $\begin{aligned} & 7.9 \\ & 8.7 \end{aligned}$ | ns |
| $t_{\text {PWH }}$ | Minimum Pulse Width High | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ | $\begin{aligned} & 8.0 \\ & 8.4 \end{aligned}$ |  | ns |
| $t_{\text {PWL }}$ | Minimum Pulse Width Low | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ | $\begin{aligned} & 1.5 \\ & 2.2 \end{aligned}$ |  | ns |
| $t_{\text {CKSW }}$ | Maximum Skew | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ |  | $\begin{aligned} & 1.5 \\ & 2.3 \end{aligned}$ | ns |
| $t_{p}$ | Minimum Period | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ | $\begin{aligned} & 16.3 \\ & 17.5 \end{aligned}$ |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Maximum Frequency | $\begin{aligned} & \mathrm{FO}=16 \\ & \mathrm{FO}=128 \end{aligned}$ |  | $\begin{aligned} & 60 \\ & 50 \end{aligned}$ | MHz |

Note:

1. These parameters should be used for estimating device performance. Optimization techniques may further reduce delays by 0 to 4 ns . Routing delays are for typical designs across worst-case operating conditions. Post-route timing analysis or simulation is required to determineactual worst-case performance. Post-routetiming is based on actual routing delay measurements performed on the device prior to shipment.

RT1020, A1020B Timing Characteristics (continued)
(Worst-Case Military Conditions)

| Output Module Timing |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Units |
| TTL Output Module Timing ${ }^{1}$ |  |  |  |  |
| $t_{\text {DLH }}$ <br> $t_{\text {DHL }}$ <br> $t_{\text {ENZH }}$ <br> tenzl <br> tenhz <br> tenlz <br> $\mathrm{d}_{\text {TLH }}$ <br> $\mathrm{d}_{\mathrm{THL}}$ | Data to Pad High <br> Data to Pad Low <br> Enable Pad Z to High <br> Enable Pad $Z$ to Low <br> Enable Pad High to $Z$ <br> Enable Pad Low to Z <br> Delta Low to High <br> Delta High to Low |  | $\begin{gathered} \hline 8.3 \\ 9.3 \\ 8.1 \\ 9.8 \\ 12.3 \\ 11.1 \\ 0.07 \\ 0.10 \end{gathered}$ | ns <br> ns <br> ns <br> ns <br> ns <br> ns <br> ns/pF <br> ns/pF |
| CMOS Output Module Timing ${ }^{1}$ |  |  |  |  |
| $t_{\text {DLH }}$ <br> $t_{\text {DHL }}$ <br> $t_{\text {ENZH }}$ <br> tenzl <br> tenhz <br> tenlz <br> $\mathrm{d}_{\text {TLH }}$ <br> $\mathrm{d}_{\mathrm{THL}}$ | Data to Pad High <br> Data to Pad Low <br> Enable Pad Z to High <br> Enable Pad Z to Low <br> Enable Pad High to $Z$ <br> Enable Pad Low to Z <br> Delta Low to High <br> Delta High to Low |  | $\begin{gathered} \hline 9.8 \\ 7.9 \\ 7.4 \\ 10.2 \\ 12.3 \\ 11.1 \\ 0.13 \\ 0.07 \end{gathered}$ | ns ns ns ns ns ns ns/pF ns/pF |

Notes:

1. Delays based on 35 pF loading.
2. SSO information can be found in the "Simultaneous Switching Output Limits for Actel FPGAs" Application Note in the 1996 Actel Data Book.

## RT1280A, A1280A Timing Characteristics

(Worst-Case Military Conditions)

| Logic Module Propagation Delays ${ }^{1}$ |  | '-1 Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {PD1 }}$ | Single Module |  | 5.2 |  | 6.1 | ns |
| $\mathrm{t}_{\mathrm{CO}}$ | Sequential Clock-to-Q |  | 5.2 |  | 6.1 | ns |
| $\mathrm{t}_{\mathrm{GO}}$ | Latch G-to-Q |  | 5.2 |  | 6.1 | ns |
| $\mathrm{t}_{\mathrm{RS}}$ | Flip-Flop (Latch) Reset-to-Q |  | 5.2 |  | 6.1 | ns |
| Predicted Routing Delays ${ }^{2}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {RD1 }}$ | FO=1 Routing Delay |  | 2.4 |  | 2.8 | ns |
| $\mathrm{t}_{\mathrm{RD} 2}$ | FO=2 Routing Delay |  | 3.4 |  | 4.0 | ns |
| $\mathrm{t}_{\mathrm{RD} 3}$ | FO=3 Routing Delay |  | 4.2 |  | 4.9 | ns |
| $\mathrm{t}_{\mathrm{RD} 4}$ | FO=4 Routing Delay |  | 5.1 |  | 6.0 | ns |
| $\mathrm{t}_{\mathrm{RD} 8}$ | FO=8 Routing Delay |  | 9.2 |  | 10.8 | ns |
| Sequential Timing Characteristics ${ }^{3,4}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {SUD }}$ | Flip-Flop (Latch) Data Input Set-Up | 0.5 |  | 0.5 |  | ns |
| $\mathrm{t}_{\mathrm{HD}}$ | Flip-Flop (Latch) Data Input Hold | 0.0 |  | 0.0 |  | ns |
| tsuena | Flip-Flop (Latch) Enable Set-Up | 1.3 |  | 1.3 |  | ns |
| $\mathrm{t}_{\text {HENA }}$ | Flip-Flop (Latch) Enable Hold | 0.0 |  | 0.0 |  | ns |
| ${ }^{\text {W WCLKA }}$ | Flip-Flop (Latch) Clock Active Pulse Width | 7.4 |  | 8.6 |  | ns |
| twasyn | Flip-Flop (Latch) Asynchronous Pulse Width | 7.4 |  | 8.6 |  | ns |
| $\mathrm{t}_{\mathrm{A}}$ | Flip-Flop Clock Input Period | 16.4 |  | 22.1 |  | ns |
| $\mathrm{t}_{\mathrm{INH}}$ | Input Buffer Latch Hold | 2.5 |  | 2.5 |  | ns |
| tinsu | Input Buffer Latch Set-Up | 3.5 |  | 3.5 |  | ns |
| touth | Output Buffer Latch Hold | 0.0 |  | 0.0 |  | ns |
| toutsu | Output Buffer Latch Set-Up | 0.5 |  | 0.5 |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Flip-Flop (Latch) Clock Frequency |  | 60 |  | 41 | MHz |

## Notes:

1. For dual-module macros, use $t_{P D 1}+t_{R D 1}+t_{P D n}, t_{C O}+t_{R D 1}+t_{P D n}$, or $t_{P D 1}+t_{R D 1}+t_{S U D}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-routetiming analysis or simulation is required to determine actual worst-case performance. Post-routetiming is based on actual routing delay measurements performed on the device prior to shi pment.
3. Data applies to macros based on the S-module. Timing parameters for sequential macros constructed from C-modules can be obtai ned from the DirectTimeAnalyzer utility.
4. Set-Up and hold timing parameters for the input buffer latch are defined with respect to the PAD and the D input. External set-up/hold timing parameters must account for delay from an external PAD signal to the $G$ inputs. Delay from an external PAD signal to the G input subtracts (adds) to the internal set-up (hold) time.

## RT1280A, A1280A Timing Characteristics (continued)

(Worst-Case Military Conditions)

| Input Module Propagation Delays |  |  | '-1' Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description |  | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\mathrm{INYH}}$ | Pad-to-Y HIGH |  |  | 4.0 |  | 4.7 | ns |
| $\mathrm{t}_{\mathrm{INYL}}$ | Pad-to-Y LOW |  |  | 3.6 |  | 4.3 | ns |
| $\mathrm{t}_{\text {INGH }}$ | G-to-Y HIGH |  |  | 6.9 |  | 8.1 | ns |
| $t_{\text {INGL }}$ | G-to-Y LOW |  |  | 6.6 |  | 7.7 | ns |
| Input Module Predicted Routing Delays ${ }^{1}$ |  |  |  |  |  |  |  |
| tIRD1 | FO=1 Routing Delay |  |  | 6.2 |  | 7.3 | ns |
| tIRD2 | FO=2 Routing Delay |  |  | 7.2 |  | 8.4 | ns |
| $\mathrm{t}_{\text {IRD3 }}$ | FO=3 Routing Delay |  |  | 7.7 |  | 9.1 | ns |
| $\mathrm{t}_{\text {IRD4 }}$ | $\mathrm{FO}=4$ Routing Delay |  |  | 8.9 |  | 10.5 | ns |
| tIRD8 | FO=8 Routing Delay |  |  | 12.9 |  | 15.2 | ns |
| Global Clock Network |  |  |  |  |  |  |  |
| $t_{\text {CKH }}$ | Input LOW to HIGH | $\begin{aligned} & \hline \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ |  | $\begin{aligned} & \hline 13.3 \\ & 17.9 \end{aligned}$ |  | $\begin{aligned} & \hline 15.7 \\ & 21.1 \end{aligned}$ | ns |
| ${ }^{\text {cheL }}$ | Input HIGH to LOW | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ |  | $\begin{aligned} & 13.3 \\ & 18.2 \end{aligned}$ |  | $\begin{aligned} & 15.7 \\ & 21.4 \end{aligned}$ | ns |
| $t_{\text {PWH }}$ | Minimum Pulse Width HIGH | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ | $\begin{aligned} & 6.9 \\ & 7.9 \end{aligned}$ |  | $\begin{aligned} & 8.1 \\ & 9.3 \end{aligned}$ |  | ns |
| $t_{\text {PWL }}$ | Minimum Pulse Width LOW | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ | $\begin{aligned} & 6.9 \\ & 7.9 \end{aligned}$ |  | $\begin{aligned} & 8.1 \\ & 9.3 \end{aligned}$ |  | ns |
| $\mathrm{t}_{\text {CKSW }}$ | Maximum Skew | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ |  | $\begin{aligned} & 0.6 \\ & 3.1 \end{aligned}$ |  | $\begin{aligned} & 0.6 \\ & 3.1 \end{aligned}$ | ns |
| $\mathrm{t}_{\text {SUEXT }}$ | Input Latch External Set-Up | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ |  | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ |  | ns |
| $\mathrm{t}_{\text {HEXT }}$ | Input Latch External Hold | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ | $\begin{gathered} 8.6 \\ 13.8 \end{gathered}$ |  | $\begin{gathered} 8.6 \\ 13.8 \end{gathered}$ |  | ns |
| $t_{p}$ | Minimum Period | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ | $\begin{aligned} & 13.7 \\ & 16.0 \end{aligned}$ |  | $\begin{aligned} & 16.2 \\ & 18.9 \end{aligned}$ |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Maximum Frequency | $\begin{aligned} & \mathrm{FO}=32 \\ & \mathrm{FO}=384 \end{aligned}$ |  | $\begin{aligned} & 73 \\ & 63 \end{aligned}$ |  | $\begin{aligned} & 62 \\ & 53 \end{aligned}$ | MHz |

## Note:

1. These parameters should be used for estimating device performance. Optimization techniques may further reduce delays by 0 to 4 ns . Routing delays are for typical designs across worst-case operating conditions. Post-route timing analysis or simulation is required to determineactual worst-case performance. Post-routetiming is based on actual routing delay measurements performed on thedevi ceprior to shipment.

RT1280A, A1280A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| Output Module Timing |  | '-1 Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min | Max. | Min. | Max. | Units |
| TTL Output Module Timing ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{DLH}}$ <br> $t_{\text {DHL }}$ <br> $t_{\text {ENZH }}$ <br> $t_{\text {ENZL }}$ <br> tenHz <br> $t_{\text {ENLZ }}$ <br> $t_{G L H}$ <br> $\mathrm{t}_{\mathrm{GHL}}$ <br> $\mathrm{d}_{\mathrm{TLH}}$ <br> $d_{\text {THL }}$ | Data-to-Pad HIGH <br> Data-to-Pad LOW <br> Enable-to-Pad Z to HIGH <br> Enable-to-Pad Z to LOW <br> Enable-to-Pad HIGH to Z <br> Enable-to-Pad LOW to Z <br> G-to-Pad HIGH <br> G-to-Pad LOW <br> Delta LOW to HIGH <br> Delta HIGH to LOW |  | $\begin{gathered} \hline 11.0 \\ 13.9 \\ 12.3 \\ 16.1 \\ 9.8 \\ 11.5 \\ 12.4 \\ 15.5 \\ 0.09 \\ 0.17 \end{gathered}$ |  | $\begin{aligned} & \hline 13.0 \\ & 16.4 \\ & 14.4 \\ & 19.0 \\ & 11.5 \\ & 13.6 \\ & 14.6 \\ & 18.2 \\ & 0.11 \\ & 0.20 \end{aligned}$ | ns <br> ns <br> ns <br> ns <br> ns <br> ns <br> ns <br> ns <br> ns/pF <br> ns/pF |
| CMOS Output Module Timing ${ }^{1}$ |  |  |  |  |  |  |
| $t_{D L H}$ <br> $\mathrm{t}_{\mathrm{DHL}}$ <br> $t_{\text {ENZH }}$ <br> $t_{\text {ENZL }}$ <br> tenhz <br> $t_{\text {ENLZ }}$ <br> $t_{G L H}$ <br> $\mathrm{t}_{\mathrm{GHL}}$ <br> $\mathrm{d}_{\mathrm{TLH}}$ <br> $\mathrm{d}_{\text {THL }}$ | Data-to-Pad HIGH <br> Data-to-Pad LOW <br> Enable-to-Pad Z to HIGH <br> Enable-to-Pad Z to LOW <br> Enable-to-Pad HIGH to Z <br> Enable-to-Pad LOW to Z <br> G-to-Pad HIGH <br> G-to-Pad LOW <br> Delta LOW to HIGH <br> Delta HIGH to LOW |  | $\begin{gathered} \hline 14.0 \\ 11.7 \\ 12.3 \\ 16.1 \\ 9.8 \\ 11.5 \\ 12.4 \\ 15.5 \\ 0.17 \\ 0.12 \end{gathered}$ |  | $\begin{aligned} & 16.5 \\ & 13.7 \\ & 14.4 \\ & 19.0 \\ & 11.5 \\ & 13.6 \\ & 14.6 \\ & 18.2 \\ & 0.20 \\ & 0.15 \end{aligned}$ | ns ns ns ns ns ns ns ns ns/pF ns/pF |

Notes:

1. Delays based on 50 pF loading.
2. SSO information can be found in the "Si multaneously Switching Output Limits for Actel FPGAs" application note.

## RT1425A, A1425A Timing Characteristics

(Worst-Case Military Conditions)

| Logic Module Propagation Delays ${ }^{1}$ |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {PD }}$ | Internal Array Module |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\mathrm{CO}}$ | Sequential Clock to Q |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\text {CLR }}$ | Asynchronous Clear to Q |  | 3.0 |  | 3.5 | ns |
| Predicted Routing Delays ${ }^{2}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {RD1 }}$ | FO=1 Routing Delay |  | 1.3 |  | 1.5 | ns |
| $\mathrm{t}_{\mathrm{RD} 2}$ | FO=2 Routing Delay |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\text {RD3 }}$ | FO=3 Routing Delay |  | 2.1 |  | 2.5 | ns |
| $\mathrm{t}_{\text {RD4 }}$ | FO=4 Routing Delay |  | 2.6 |  | 2.9 | ns |
| $\mathrm{t}_{\text {RD8 }}$ | FO=8 Routing Delay |  | 4.2 |  | 4.9 | ns |
| Logic Module Sequential Timing |  |  |  |  |  |  |
| $\mathrm{t}_{\text {SUD }}$ | Flip-Flop (Latch) Data Input Setup | 0.9 |  | 1.0 |  | ns |
| $\mathrm{t}_{\mathrm{HD}}$ | Flip-Flop (Latch) Data Input Hold | 0.0 |  | 0.0 |  | ns |
| tsuena | Flip-Flop (Latch) Enable Setup | 0.9 |  | 1.0 |  | ns |
| $t_{\text {HENA }}$ | Flip-Flop (Latch) Enable Hold | 0.0 |  | 0.0 |  | ns |
| twasyn | Asynchronous Pulse Width | 3.8 |  | 4.4 |  | ns |
| twCLKA | Flip-Flop Clock Pulse Width | 3.8 |  | 4.4 |  | ns |
| $t_{\text {A }}$ | Flip-Flop Clock Input Period | 7.9 |  | 9.3 |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Flip-Flop Clock Frequency |  | 125 |  | 100 | MHz |

## Notes:

1. For dual-module macros, use $t_{P D}+t_{R D 1}+t_{P D n}, t_{C O}+t_{R D 1}+t_{P D n}$, or $t_{P D 1}+t_{R D 1}+t_{S U D}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-routetiming analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shi pment.

RT1425A, A1425A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| I/O Module Input Propagation Delays |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {INY }}$ | Input Data Pad to Y |  | 4.2 |  | 4.9 | ns |
| ticky | Input Reg IOCLK Pad to $Y$ |  | 7.0 |  | 8.2 | ns |
| tocky | Output Reg IOCLK Pad to Y |  | 7.0 |  | 8.2 | ns |
| ticlir | Input Asynchronous Clear to $Y$ |  | 7.0 |  | 8.2 | ns |
| tocliry | Output Asynchronous Clear to Y |  | 7.0 |  | 8.2 | ns |
| Predicted Input Routing Delays ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {IRD1 }}$ | FO=1 Routing Delay |  | 1.3 |  | 1.5 | ns |
| $\mathrm{t}_{\text {IRD2 }}$ | FO=2 Routing Delay |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\text {IRD3 }}$ | FO=3 Routing Delay |  | 2.1 |  | 2.5 | ns |
| $\mathrm{t}_{\text {RD4 }}$ | FO=4 Routing Delay |  | 2.6 |  | 2.9 | ns |
| $\mathrm{t}_{\text {IRD8 }}$ | FO=8 Routing Delay |  | 4.2 |  | 4.9 | ns |
| I/O Module Sequential Timing |  |  |  |  |  |  |
| tinh | Input F-F Data Hold (w.r.t. IOCLK Pad) | 0.0 |  | 0.0 |  | ns |
| $\mathrm{t}_{\text {INSU }}$ | Input F-F Data Setup (w.r.t. IOCLK Pad) | 2.1 |  | 2.4 |  | ns |
| $\mathrm{t}_{\text {IDEH }}$ | Input Data Enable Hold (w.r.t. IOCLK Pad) | 0.0 |  | 0.0 |  | ns |
| tIDESU | Input Data Enable Setup (w.r.t. IOCLK Pad) | 8.7 |  | 10.0 |  | ns |
| touth | Output F-F Data Hold (w.r.t. IOCLK Pad) | 1.1 |  | 1.2 |  | ns |
| toutsu | Output F-F Data Setup (w.r.t. IOCLK Pad) | 1.1 |  | 1.2 |  | ns |
| $t_{\text {ODEH }}$ | Output Data Enable Hold (w.r.t. IOCLK Pad) | 0.5 |  | 0.6 |  | ns |
| todesu | Output Data Enable Setup (w.r.t. IOCLK Pad) | 2.0 |  | 2.4 |  | ns |

## Note:

1. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

## RT1425A, A1425A Timing Characteristics (continued)

(Worst-Case Military Conditions)

| I/O Module - TTL Output Timing ${ }^{1}$ |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min | Max. | Min | Max. | Units |
| $\mathrm{t}_{\text {DHS }}$ | Data to Pad, High Slew |  | 7.5 |  | 8.9 | ns |
| tbls | Data to Pad, Low Slew |  | 11.9 |  | 14.0 | ns |
| $t_{\text {ENZHS }}$ | Enable to Pad, Z to H/L, Hi Slew |  | 6.0 |  | 7.0 | ns |
| tenzLs | Enable to Pad, Z to H/L, Lo Slew |  | 10.9 |  | 12.8 | ns |
| $t_{\text {ENHSZ }}$ | Enable to Pad, H/L to Z, Hi Slew |  | 9.9 |  | 11.6 | ns |
| tenlsz | Enable to Pad, H/L to Z, Lo Slew |  | 9.9 |  | 11.6 | ns |
| $\mathrm{t}_{\text {CKHS }}$ | IOCLK Pad to Pad H/L, Hi Slew |  | 10.5 |  | 11.6 | ns |
| $\mathrm{t}_{\text {CKLS }}$ | IOCLK Pad to Pad H/L, Lo Slew |  | 15.7 |  | 17.4 | ns |
| $\mathrm{d}_{\text {TLHHS }}$ | Delta Low to High, Hi Slew |  | 0.04 |  | 0.04 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {TLHLS }}$ | Delta Low to High, Lo Slew |  | 0.07 |  | 0.08 | ns/pF |
| $\mathrm{d}_{\text {THLHS }}$ | Delta High to Low, Hi Slew |  | 0.05 |  | 0.06 | ns/pF |
| $\mathrm{d}_{\text {THLLS }}$ | Delta High to Low, Lo Slew |  | 0.07 |  | 0.08 | ns/pF |
| I/O Module - CMOS Output Timing ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {DHS }}$ | Data to Pad, High Slew |  | 9.2 |  | 10.8 | ns |
| $t_{\text {bLs }}$ | Data to Pad, Low Slew |  | 17.3 |  | 20.3 | ns |
| $t_{\text {ENZHS }}$ | Enable to Pad, Z to H/L, Hi Slew |  | 7.7 |  | 9.1 | ns |
| tenzls | Enable to Pad, Z to H/L, Lo Slew |  | 13.1 |  | 15.5 | ns |
| $t_{\text {ENHSZ }}$ | Enable to Pad, H/L to Z, Hi Slew |  | 9.9 |  | 11.6 | ns |
| tenlsz | Enable to Pad, H/L to Z, Lo Slew |  | 10.5 |  | 11.6 | ns |
| $\mathrm{t}_{\text {CKHS }}$ | IOCLK Pad to Pad H/L, Hi Slew |  | 12.5 |  | 13.7 | ns |
| ${ }_{\text {t CKLS }}$ | IOCLK Pad to Pad H/L, Lo Slew |  | 18.1 |  | 20.1 | ns |
| $\mathrm{d}_{\text {TLHHS }}$ | Delta Low to High, Hi Slew |  | 0.06 |  | 0.07 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {TLHLS }}$ | Delta Low to High, Lo Slew |  | 0.11 |  | 0.13 | ns/pF |
| $\mathrm{d}_{\text {THLHS }}$ | Delta High to Low, Hi Slew |  | 0.04 |  | 0.05 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {THLLS }}$ | Delta High to Low, Lo Slew |  | 0.05 |  | 0.06 | ns/pF |

## Notes:

1. Delays based on 35 pF loading.
2. SSO information can be found in the "Si multaneously Switching Output Limits for Actel FPGAs" application note.

RT1425A, A1425A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| Dedicated (Hard-Wired) I/O Clock Network |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {IOCKH }}$ | Input Low to High (Pad to I/O Module Input) |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\text {IOPWH }}$ | Minimum Pulse Width High | 3.9 |  | 4.4 |  | ns |
| $\mathrm{t}_{\text {IOPWL }}$ | Minimum Pulse Width Low | 3.9 |  | 4.4 |  | ns |
| tiosarw | Minimum Asynchronous Pulse Width | 3.9 |  | 4.4 |  | ns |
| tiocksw | Maximum Skew |  | 0.5 |  | 0.5 | ns |
| $\mathrm{t}_{\mathrm{IOP}}$ | Minimum Period | 7.9 |  | 9.3 |  | ns |
| fiomax | Maximum Frequency |  | 125 |  | 100 | MHz |
| Dedicated (Hard-Wired) Array Clock Network |  |  |  |  |  |  |
| thekh $^{\text {l }}$ | Input Low to High <br> (Pad to S-Module Input) |  | 4.6 |  | 5.3 | ns |
| $\mathrm{t}_{\mathrm{HCKL}}$ | Input High to Low <br> (Pad to S-Module Input) |  | 4.6 |  | 5.3 | ns |
| $\mathrm{t}_{\text {HPWH }}$ | Minimum Pulse Width High | 3.9 |  | 4.4 |  | ns |
| $t_{\text {HPWL }}$ | Minimum Pulse Width Low | 3.9 |  | 4.4 |  | ns |
| $\mathrm{t}_{\text {HCKSW }}$ | Maximum Skew |  | 0.4 |  | 0.4 | ns |
| $t_{\text {HP }}$ | Minimum Period | 7.9 |  | 9.3 |  | ns |
| $\mathrm{f}_{\text {HMAX }}$ | Maximum Frequency |  | 125 |  | 100 | MHz |
| Routed Array Clock Networks |  |  |  |  |  |  |
| $\mathrm{t}_{\text {RCKH }}$ | Input Low to High (FO=64) |  | 5.5 |  | 6.4 | ns |
| $t_{\text {RCKL }}$ | Input High to Low (FO=64) |  | 6.0 |  | 7.0 | ns |
| $\mathrm{t}_{\text {RPWH }}$ | Min. Pulse Width High (FO=64) | 4.9 |  | 5.7 |  | ns |
| $\mathrm{t}_{\text {RPWL }}$ | Min. Pulse Width Low (FO=64) | 4.9 |  | 5.7 |  | ns |
| $t_{\text {RCKSW }}$ | Maximum Skew ( $\mathrm{FO}=128$ ) |  | 1.1 |  | 1.2 | ns |
| $t_{\text {RP }}$ | Minimum Period (FO=64) | 10.1 |  | 11.6 |  | ns |
| $\mathrm{f}_{\text {RMAX }}$ | Maximum Frequency (FO=64) |  | 100 |  | 85 | MHz |
| Clock-to-Clock Skews |  |  |  |  |  |  |
| $\mathrm{t}_{\text {IOHCKSW }}$ | I/O Clock to H-Clock Skew | 0.0 | 3.0 | 0.0 | 3.0 | ns |
| tiorcksw | I/O Clock to R-Clock Skew | 0.0 | 3.0 | 0.0 | 3.0 | ns |
| thrcksw | H-Clock to R-Clock Skew $\begin{aligned} & (F O=64) \\ & (F O=50 \% \text { max. }) \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \end{aligned}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \end{aligned}$ |

## Note:

1. Delays based on 35 pF loading.

## RT1460A, A1460A Timing Characteristics

(Worst-Case Military Conditions)

| Logic Module Propagation Delays ${ }^{1}$ |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $t_{\text {PD }}$ | Internal Array Module |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\mathrm{CO}}$ | Sequential Clock to Q |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\text {CLR }}$ | Asynchronous Clear to Q |  | 3.0 |  | 3.5 | ns |
| Predicted Routing Delays ${ }^{2}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {RD1 }}$ | FO=1 Routing Delay |  | 1.3 |  | 1.5 | ns |
| $\mathrm{t}_{\mathrm{RD} 2}$ | FO=2 Routing Delay |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\text {RD3 }}$ | FO=3 Routing Delay |  | 2.1 |  | 2.5 | ns |
| $\mathrm{t}_{\text {RD4 }}$ | FO=4 Routing Delay |  | 2.6 |  | 2.9 | ns |
| $\mathrm{t}_{\text {RD8 }}$ | FO=8 Routing Delay |  | 4.2 |  | 4.9 | ns |
| Logic Module Sequential Timing |  |  |  |  |  |  |
| $\mathrm{t}_{\text {SUD }}$ | Flip-Flop (Latch) Data Input Setup | 0.9 |  | 1.0 |  | ns |
| $t_{\text {HD }}$ | Flip-Flop (Latch) Data Input Hold | 0.0 |  | 0.0 |  | ns |
| tsuena | Flip-Flop (Latch) Enable Setup | 0.9 |  | 1.0 |  | ns |
| $t_{\text {HENA }}$ | Flip-Flop (Latch) Enable Hold | 0.0 |  | 0.0 |  | ns |
| $\mathrm{t}_{\text {WASYN }}$ | Asynchronous Pulse Width | 4.8 |  | 5.6 |  | ns |
| $\mathrm{t}_{\text {WCLKA }}$ | Flip-Flop Clock Pulse Width | 4.8 |  | 5.6 |  | ns |
| $\mathrm{t}_{\mathrm{A}}$ | Flip-Flop Clock Input Period | 9.9 |  | 11.6 |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Flip-Flop Clock Frequency |  | 100 |  | 85 | MHz |

## Notes:

1. For dual-module macros, use $t_{P D}+t_{R D 1}+t_{P D n}, t_{C O}+t_{R D 1}+t_{P D n}$, or $t_{P D 1}+t_{R D 1}+t_{S U D}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-routetiming analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

RT1460A, A1460A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| I/O Module Input Propagation Delays |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {INY }}$ | Input Data Pad to Y |  | 4.2 |  | 4.9 | ns |
| tICKY | Input Reg IOCLK Pad to Y |  | 7.0 |  | 8.2 | ns |
| tocky | Output Reg IOCLK Pad to Y |  | 7.0 |  | 8.2 | ns |
| $\mathrm{t}_{\text {ICLRY }}$ | Input Asynchronous Clear to Y |  | 7.0 |  | 8.2 | ns |
| toclRy | Output Asynchronous Clear to Y |  | 7.0 |  | 8.2 | ns |
| Predicted Input Routing Delays ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {IRD1 }}$ | FO=1 Routing Delay |  | 1.3 |  | 1.5 | ns |
| $\mathrm{t}_{\text {IRD2 }}$ | FO=2 Routing Delay |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\text {IRD3 }}$ | FO=3 Routing Delay |  | 2.1 |  | 2.5 | ns |
| $\mathrm{t}_{\text {IRD4 }}$ | FO=4 Routing Delay |  | 2.6 |  | 2.9 | ns |
| $\mathrm{t}_{\text {IRD8 }}$ | FO=8 Routing Delay |  | 4.2 |  | 4.9 | ns |
| I/O Module Sequential Timing |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{INH}}$ | Input F-F Data Hold (w.r.t. IOCLK Pad) | 0.0 |  | 0.0 |  | ns |
| $\mathrm{t}_{\text {INSU }}$ | Input F-F Data Setup (w.r.t. IOCLK Pad) | 2.1 |  | 2.4 |  | ns |
| $t_{\text {IDEH }}$ | Input Data Enable Hold (w.r.t. IOCLK Pad) | 0.0 |  | 0.0 |  | ns |
| $\mathrm{t}_{\text {IDESU }}$ | Input Data Enable Setup (w.r.t. IOCLK Pad) | 8.7 |  | 10.0 |  | ns |
| touth | Output F-F Data Hold (w.r.t. IOCLK Pad) | 1.1 |  | 1.2 |  | ns |
| toutsu | Output F-F Data Setup (w.r.t. IOCLK Pad) | 1.1 |  | 1.2 |  | ns |
| $t_{\text {ODEH }}$ | Output Data Enable Hold (w.r.t. IOCLK Pad) | 0.5 |  | 0.6 |  | ns |
| $\mathrm{t}_{\text {ODESU }}$ | Output Data Enable Setup (w.r.t. IOCLK Pad) | 2.0 |  | 2.4 |  | ns |

Note:

1. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-routetiming analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shi pment.

## RT1460A, A1460A Timing Characteristics (continued)

(Worst-Case Military Conditions)

| I/O Module - TTL Output Timing ${ }^{1}$ |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {DHS }}$ | Data to Pad, High Slew |  | 7.5 |  | 8.9 | ns |
| tbls | Data to Pad, Low Slew |  | 11.9 |  | 14.0 | ns |
| $t_{\text {ENZHS }}$ | Enable to Pad, Z to $\mathrm{H} / \mathrm{L}$, Hi Slew |  | 6.0 |  | 7.0 | ns |
| $t_{\text {ENZLS }}$ | Enable to Pad, Z to H/L, Lo Slew |  | 10.9 |  | 12.8 | ns |
| $t_{\text {ENHSZ }}$ | Enable to Pad, H/L to Z, Hi Slew |  | 11.5 |  | 13.5 | ns |
| $\mathrm{t}_{\text {ENLSZ }}$ | Enable to Pad, H/L to Z, Lo Slew |  | 10.9 |  | 12.8 | ns |
| $\mathrm{t}_{\text {CKHS }}$ | IOCLK Pad to Pad H/L, Hi Slew |  | 11.6 |  | 13.4 | ns |
| $\mathrm{t}_{\text {CKLS }}$ | IOCLK Pad to Pad H/L, Lo Slew |  | 17.8 |  | 19.8 | ns |
| dtLhHS | Delta Low to High, Hi Slew |  | 0.04 |  | 0.04 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {TLHLS }}$ | Delta Low to High, Lo Slew |  | 0.07 |  | 0.08 | ns/pF |
| $\mathrm{d}_{\text {THLHS }}$ | Delta High to Low, Hi Slew |  | 0.05 |  | 0.06 | ns/pF |
| $\mathrm{d}_{\text {THLLS }}$ | Delta High to Low, Lo Slew |  | 0.07 |  | 0.08 | ns/pF |
| I/O Module - CMOS Output Timing ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {DHS }}$ | Data to Pad, High Slew |  | 9.2 |  | 10.8 | ns |
| tbls | Data to Pad, Low Slew |  | 17.3 |  | 20.3 | ns |
| $t_{\text {ENZHS }}$ | Enable to Pad, Z to $\mathrm{H} / \mathrm{L}$, Hi Slew |  | 7.7 |  | 9.1 | ns |
| $t_{\text {ENZLS }}$ | Enable to Pad, Z to H/L, Lo Slew |  | 13.1 |  | 15.5 | ns |
| $t_{\text {ENHSZ }}$ | Enable to Pad, H/L to Z, Hi Slew |  | 10.9 |  | 12.8 | ns |
| $t_{\text {ENLS }}$ | Enable to Pad, H/L to Z, Lo Slew |  | 10.9 |  | 12.8 | ns |
| $\mathrm{t}_{\text {CKHS }}$ | IOCLK Pad to Pad H/L, Hi Slew |  | 14.1 |  | 16.0 | ns |
| ${ }_{\text {t }}^{\text {CKLS }}$ | IOCLK Pad to Pad H/L, Lo Slew |  | 20.2 |  | 22.4 | ns |
| $\mathrm{d}_{\text {TLHHS }}$ | Delta Low to High, Hi Slew |  | 0.06 |  | 0.07 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {TLHLS }}$ | Delta Low to High, Lo Slew |  | 0.11 |  | 0.13 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {THLHS }}$ | Delta High to Low, Hi Slew |  | 0.04 |  | 0.05 | ns/pF |
| $\mathrm{d}_{\text {THLLS }}$ | Delta High to Low, Lo Slew |  | 0.05 |  | 0.06 | ns/pF |

## Notes:

1. Delays based on 35 pF loading.
2. SSO information can be found in the "Si multaneously Switching Output Limits for Actel FPGAs" application note.

RT1460A, A1460A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| Dedicated (Hard-Wired) I/O Clock Network |  | -1 Speed |  | Std Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {IOCKH }}$ | Input Low to High <br> (Pad to I/O Module Input) |  | 3.5 |  | 4.1 | ns |
| $\mathrm{t}_{\text {IOPWH }}$ | Minimum Pulse Width High | 4.8 |  | 5.7 |  | ns |
| $\mathrm{t}_{\text {IOPWL }}$ | Minimum Pulse Width Low | 4.8 |  | 5.7 |  | ns |
| tIoSAPW | Minimum Asynchronous Pulse Width | 3.9 |  | 4.4 |  | ns |
| tiocksw | Maximum Skew |  | 0.9 |  | 1.0 | ns |
| $\mathrm{t}_{\mathrm{IOP}}$ | Minimum Period | 9.9 |  | 11.6 |  | ns |
| flomax | Maximum Frequency |  | 100 |  | 85 | MHz |
| Dedicated (Hard-Wired) Array Clock Network |  |  |  |  |  |  |
| $\mathrm{t}_{\text {HCKH }}$ | Input Low to High <br> (Pad to S-Module Input) |  | 5.5 |  | 6.4 | ns |
| $\mathrm{t}_{\text {HCKL }}$ | Input High to Low <br> (Pad to S-Module Input) |  | 5.5 |  | 6.4 | ns |
| $\mathrm{t}_{\text {HPW }}$ | Minimum Pulse Width High | 4.8 |  | 5.7 |  | ns |
| $t_{\text {HPWL }}$ | Minimum Pulse Width Low | 4.8 |  | 5.7 |  | ns |
| $t_{\text {HCKSW }}$ | Maximum Skew |  | 0.9 |  | 1.0 | ns |
| $t_{\text {HP }}$ | Minimum Period | 9.9 |  | 11.6 |  | ns |
| $\mathrm{f}_{\text {HMAX }}$ | Maximum Frequency |  | 100 |  | 85 | MHz |
| Routed Array Clock Networks |  |  |  |  |  |  |
| $\mathrm{t}_{\text {RCKH }}$ | Input Low to High (FO=256) |  | 9.0 |  | 10.5 | ns |
| $\mathrm{t}_{\text {RCKL }}$ | Input High to Low (FO=256) |  | 9.0 |  | 10.5 | ns |
| $t_{\text {RPWH }}$ | Min. Pulse Width High (FO=256) | 6.3 |  | 7.1 |  | ns |
| $t_{\text {RPWL }}$ | Min. Pulse Width Low (FO=256) | 6.3 |  | 7.1 |  | ns |
| $t_{\text {RCKSW }}$ | Maximum Skew ( $\mathrm{FO}=128$ ) |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\mathrm{RP}}$ | Minimum Period ( $\mathrm{FO}=256$ ) | 12.9 |  | 14.5 |  | ns |
| $\mathrm{f}_{\text {RMAX }}$ | Maximum Frequency ( $\mathrm{FO}=256$ ) |  | 75 |  | 65 | MHz |
| Clock-to-Clock Skews |  |  |  |  |  |  |
| tiohcksw | I/O Clock to H-Clock Skew | 0.0 | 3.0 | 0.0 | 3.0 | ns |
| tiorcksw | I/O Clock to R-Clock Skew | 0.0 | 5.0 | 0.0 | 5.0 | ns |
| thrcksw | H-Clock to R-Clock Skew $\begin{aligned} & (F O=64) \\ & (F O=50 \% \text { max. }) \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \end{aligned}$ | $\begin{aligned} & \mathrm{ns} \\ & \mathrm{~ns} \end{aligned}$ |

## Note:

1. Delays based on 35 pF loading.

## RT14100A, A14100A Timing Characteristics

(Worst-Case Military Conditions)

| Logic Module Propagation Delays ${ }^{1}$ |  | '-1' Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {PD }}$ | Internal Array Module |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\mathrm{CO}}$ | Sequential Clock-to-Q |  | 3.0 |  | 3.5 | ns |
| $\mathrm{t}_{\text {CLR }}$ | Asynchronous Clear-to-Q |  | 3.0 |  | 3.5 | ns |
| Predicted Routing Delays ${ }^{2}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{RD} 1}$ | FO=1 Routing Delay |  | 1.3 |  | 1.5 | ns |
| $\mathrm{t}_{\text {RD2 }}$ | FO=2 Routing Delay |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\text {RD3 }}$ | FO=3 Routing Delay |  | 2.1 |  | 2.5 | ns |
| $\mathrm{t}_{\text {RD4 }}$ | FO=4 Routing Delay |  | 2.6 |  | 2.9 | ns |
| $\mathrm{t}_{\text {RD8 }}$ | FO=8 Routing Delay |  | 4.2 |  | 4.9 | ns |
| Logic Module Sequential Timing |  |  |  |  |  |  |
| $\mathrm{t}_{\text {SUD }}$ | Flip-Flop (Latch) Data Input Set-Up | 1.0 |  | 1.0 |  | ns |
| $\mathrm{t}_{\mathrm{HD}}$ | Flip-Flop (Latch) Data Input Hold | 0.6 |  | 0.6 |  | ns |
| $\mathrm{t}_{\text {SUENA }}$ | Flip-Flop (Latch) Enable Set-Up | 1.0 |  | 1.0 |  | ns |
| $\mathrm{t}_{\text {HENA }}$ | Flip-Flop (Latch) Enable Hold | 0.6 |  | 0.6 |  | ns |
| $t_{\text {WASYN }}$ | Asynchronous Pulse Width | 4.8 |  | 5.6 |  | ns |
| $t_{\text {WCLKA }}$ | Flip-Flop Clock Pulse Width | 4.8 |  | 5.6 |  | ns |
| $\mathrm{t}_{\mathrm{A}}$ | Flip-Flop Clock Input Period | 9.9 |  | 11.6 |  | ns |
| $\mathrm{f}_{\text {MAX }}$ | Flip-Flop Clock Frequency |  | 100 |  | 85 | MHz |

## Notes:

1. For dual-module macros, use $t_{P D}+t_{R D 1}+t_{P D n}, t_{C O}+t_{R D 1}+t_{P D n}$, or $t_{P D 1}+t_{R D 1}+t_{S U D}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on thedevice prior to shipment.

RT14100A, A14100A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| I/O Module Input Propagation Delays |  | '-1 Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {INY }}$ | Input Data Pad-to-Y |  | 4.2 |  | 4.9 | ns |
| $\mathrm{t}_{\text {ICKY }}$ | Input Reg IOCLK Pad-to-Y |  | 7.0 |  | 8.2 | ns |
| tocky | Output Reg IOCLK Pad-to-Y |  | 7.0 |  | 8.2 | ns |
| tICLRY | Input Asynchronous Clear-to-Y |  | 7.0 |  | 8.2 | ns |
| toclRy | Output Asynchronous Clear-to-Y |  | 7.0 |  | 8.2 | ns |
| Predicted Input Routing Delays ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {IRD1 }}$ | FO=1 Routing Delay |  | 1.3 |  | 1.5 | ns |
| $\mathrm{t}_{\text {IRD2 }}$ | FO=2 Routing Delay |  | 1.9 |  | 2.1 | ns |
| $\mathrm{t}_{\text {IRD3 }}$ | FO=3 Routing Delay |  | 2.1 |  | 2.5 | ns |
| $\mathrm{t}_{\text {IRD4 }}$ | FO=4 Routing Delay |  | 2.6 |  | 2.9 | ns |
| $\mathrm{t}_{\text {IRD8 }}$ | FO=8 Routing Delay |  | 4.2 |  | 4.9 | ns |
| I/O Module Sequential Timing |  |  |  |  |  |  |
| $\mathrm{t}_{\text {INH }}$ | Input Flip-Flop Data Hold | 0.0 |  | 0.0 |  | ns |
| $\mathrm{t}_{\text {INSU }}$ | Input Flip-Flop Data Set-Up | 2.1 |  | 2.4 |  | ns |
| $\mathrm{t}_{\text {IDEH }}$ | Input Data Enable Hold | 0.0 |  | 0.0 |  | ns |
| $\mathrm{t}_{\text {IDESU }}$ | Input Data Enable Set-Up | 8.7 |  | 10.0 |  | ns |
| touth | Output Flip-Flop Data Hold | 1.2 |  | 1.2 |  | ns |
| toutsu | Output Flip-Flop Data Set-Up | 1.2 |  | 1.2 |  | ns |
| $\mathrm{t}_{\text {ODEH }}$ | Output Data Enable Hold | 0.6 |  | 0.6 |  | ns |
| todesu | Output Data Enable Set-Up | 2.4 |  | 2.4 |  | ns |

Note:

1. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-routetiming analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shi pment.

## RT14100A, A14100A Timing Characteristics (continued)

(Worst-Case Military Conditions)

| I/O Module - TTL Output Timing ${ }^{1}$ |  | '-1 Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {DHS }}$ | Data-to-Pad, High Slew |  | 7.5 |  | 8.9 | ns |
| $t_{\text {DLS }}$ | Data-to-Pad, Low Slew |  | 11.9 |  | 14.0 | ns |
| $t_{\text {ENZHS }}$ | Enable-to-Pad, Z to H/L, High Slew |  | 6.0 |  | 7.0 | ns |
| $t_{\text {ENZLS }}$ | Enable-to-Pad, Z to H/L, Low Slew |  | 10.9 |  | 12.8 | ns |
| $t_{\text {ENHSZ }}$ | Enable-to-Pad, H/L to Z, High Slew |  | 11.9 |  | 14.0 | ns |
| $t_{\text {ENLSZ }}$ | Enable-to-Pad, H/L to Z, Low Slew |  | 10.9 |  | 12.8 | ns |
| $\mathrm{t}_{\text {CKHS }}$ | IOCLK Pad-to-Pad H/L, High Slew |  | 12.2 |  | 14.0 | ns |
| ${ }_{\text {t }}^{\text {CKLS }}$ | IOCLK Pad-to-Pad H/L, Low Slew |  | 17.8 |  | 17.8 | ns |
| $\mathrm{d}_{\text {TLHHS }}$ | Delta LOW to HIGH, High Slew |  | 0.04 |  | 0.04 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {TLHLS }}$ | Delta LOW to HIGH, Low Slew |  | 0.07 |  | 0.08 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {THLHS }}$ | Delta HIGH to LOW, High Slew |  | 0.05 |  | 0.06 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {THLLS }}$ | Delta HIGH to LOW, Low Slew |  | 0.07 |  | 0.08 | ns/pF |
| I/O Module - CMOS Output Timing ${ }^{1}$ |  |  |  |  |  |  |
| $\mathrm{t}_{\text {DHS }}$ | Data-to-Pad, High Slew |  | 9.2 |  | 10.8 | ns |
| tols | Data-to-Pad, Low Slew |  | 17.3 |  | 20.3 | ns |
| $t_{\text {ENZHS }}$ | Enable-to-Pad, Z to H/L, High Slew |  | 7.7 |  | 9.1 | ns |
| $t_{\text {ENZLS }}$ | Enable-to-Pad, Z to H/L, Low Slew |  | 13.1 |  | 15.5 | ns |
| $t_{\text {ENHSZ }}$ | Enable-to-Pad, H/L to Z, High Slew |  | 11.6 |  | 14.0 | ns |
| $t_{\text {EnLsz }}$ | Enable-to-Pad, H/L to Z, Low Slew |  | 10.9 |  | 12.8 | ns |
| $\mathrm{t}_{\text {CKHS }}$ | IOCLK Pad-to-Pad H/L, High Slew |  | 14.4 |  | 16.0 | ns |
| ${ }_{\text {t }}^{\text {CKLS }}$ | IOCLK Pad-to-Pad H/L, Low Slew |  | 20.2 |  | 22.4 | ns |
| $\mathrm{d}_{\text {TLHHS }}$ | Delta LOW to HIGH, High Slew |  | 0.06 |  | 0.07 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {TLHLS }}$ | Delta LOW to HIGH, Low Slew |  | 0.11 |  | 0.13 | $\mathrm{ns} / \mathrm{pF}$ |
| $\mathrm{d}_{\text {THLHS }}$ | Delta HIGH to LOW, High Slew |  | 0.04 |  | 0.05 | ns/pF |
| $\mathrm{d}_{\text {THLLS }}$ | Delta HIGH to LOW, Low Slew |  | 0.05 |  | 0.06 | ns/pF |

## Notes:

1. Delays based on 35 pF loading.
2. SSO information can be found in the "Si multaneously Switching Output Limits for Actel FPGAs" application note.

RT14100A, A14100A Timing Characteristics (continued)
(Worst-Case Military Conditions)

| Dedicated (Hard-Wired) I/O Clock Network |  | '-1' Speed |  | 'Std' Speed |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Description | Min. | Max. | Min. | Max. | Units |
| $\mathrm{t}_{\text {IOCKH }}$ | Input LOW to HIGH <br> (Pad to I/O Module Input) |  | 3.5 |  | 4.1 | ns |
| $\mathrm{t}_{\text {IOPWH }}$ | Minimum Pulse Width HIGH | 4.8 |  | 5.7 |  | ns |
| $\mathrm{t}_{\text {IOPWL }}$ | Minimum Pulse Width LOW | 4.8 |  | 5.7 |  | ns |
| tiosarw | Minimum Asynchronous Pulse Width | 3.9 |  | 4.4 |  | ns |
| tiocksw | Maximum Skew |  | 0.9 |  | 1.0 | ns |
| $\mathrm{t}_{\text {IOP }}$ | Minimum Period | 9.9 |  | 11.6 |  | ns |
| fiomax | Maximum Frequency |  | 100 |  | 85 | MHz |
| Dedicated (Hard-Wired) Array Clock Network |  |  |  |  |  |  |
| $\mathrm{t}_{\mathrm{HCKH}}$ | Input LOW to HIGH <br> (Pad to S-Module Input) |  | 5.5 |  | 6.4 | ns |
| $t_{\text {HCKL }}$ | Input HIGH to LOW <br> (Pad to S-Module Input) |  | 5.5 |  | 6.4 | ns |
| $\mathrm{t}_{\text {HPW }}$ | Minimum Pulse Width HIGH | 4.8 |  | 5.7 |  | ns |
| $\mathrm{t}_{\text {HPWL }}$ | Minimum Pulse Width LOW | 4.8 |  | 5.7 |  | ns |
| $\mathrm{t}_{\text {HCKSW }}$ | Maximum Skew |  | 0.9 |  | 1.0 | ns |
| $t_{H P}$ | Minimum Period | 9.9 |  | 11.6 |  | ns |
| $\mathrm{f}_{\text {HMAX }}$ | Maximum Frequency |  | 100 |  | 85 | MHz |
| Routed Array Clock Networks |  |  |  |  |  |  |
| $t_{\text {RCKH }}$ | Input LOW to HIGH (FO=256) |  | 9.0 |  | 10.5 | ns |
| $\mathrm{t}_{\text {RCKL }}$ | Input HIGH to LOW (FO=256) |  | 9.0 |  | 10.5 | ns |
| $\mathrm{t}_{\text {RPWH }}$ | Min. Pulse Width HIGH (FO=256) | 6.3 |  | 7.1 |  | ns |
| $t_{\text {RPWL }}$ | Min. Pulse Width LOW (FO=256) | 6.3 |  | 7.1 |  | ns |
| $t_{\text {RCKSW }}$ | Maximum Skew ( $\mathrm{FO}=128$ ) |  | 1.9 |  | 2.1 | ns |
| $t_{\text {RP }}$ | Minimum Period ( $\mathrm{FO}=256$ ) | 12.9 |  | 14.5 |  | ns |
| $\mathrm{f}_{\text {RMAX }}$ | Maximum Frequency ( $\mathrm{FO}=256$ ) |  | 75 |  | 65 | MHz |
| Clock-to-Clock Skews |  |  |  |  |  |  |
| $\mathrm{t}_{\text {IOHCKSW }}$ | I/O Clock to H-Clock Skew | 0.0 | 3.5 | 0.0 | 3.5 | ns |
| tiorcksw | I/O Clock to R-Clock Skew | 0.0 | 5.0 | 0.0 | 5.0 | ns |
| thrcksw | H-Clock to R-Clock Skew $\begin{aligned} & (F O=64) \\ & (F O=50 \% \text { max. }) \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \end{aligned}$ | $\begin{aligned} & 0.0 \\ & 0.0 \end{aligned}$ | $\begin{aligned} & 1.0 \\ & 3.0 \end{aligned}$ | ns |

## Note:

1. Delays based on 35 pF loading.

## Package Pin Assignments

## 84-Pin CQFP (Top View)



| Function | RT1020, A1020B Pin Number |
| :--- | :--- |
| CLKA or I/O | 53 |
| DCLK or I/O | 62 |
| GND | $7,8,29,49,50,71$ |
| MODE | 55 |
| N/C (No Connection) | 1 |
| PRA or I/O | 63 |
| PRB or I/O | 64 |
| SDI or I/O | 61 |
| $V_{\text {CC }}$ | $14,15,22,35,56,57,77$ |

## Notes:

1. MODE should beterminated to GND through a 10K resistor to enableActionProbe usage; otherwise, it can be terminated di rectly to GND.
2. Unused I/O pins aredesignated as outputs by Designer and aredriven LOW.
3. All unassigned pins are available for useas I/Os

## Package Pin Assignments (continued)

## 132-Pin CQFP (Top View)



| Function | RT1425A, A1425A Pin Number |
| :--- | :--- |
| CLKA or I/O | 116 |
| CLKB or I/O | 117 |
| DCLK or I/O | 131 |
| GND | $2,10,26,36,42,58,65,74,90,92,101,106,122$ |
| HCLK or I/O | 50 |
| IOCLK or I/O | 98 |
| IOPCL or I/O | 64 |
| MODE | 9 |
| NC | $1,34,66,67,99,100,132$ |
| PRA or I/O | 118 |
| PRB or I/O | 48 |
| SDI or I/O | 3 |
| VCC $^{2}$ | $11,22,27,43,59,75,78,89,91,107,123$ |

## Notes:

1. Unused I/O pins aredesignated as outputs by ALS and are driven low.
2. All unassigned pins are available for useas I/Os.
3. MODE should beterminated to GND through a 10K resistor to enableActionprobe usage, other wi seit can be terminated directly to GND.

## Package Pin Assignments (continued)

## 172-Pin CQFP (Top View)



| Function | RT1280A, A1280A Pin Number |
| :--- | :--- |
| CLKA or I/O | 150 |
| CLKB or I/O | 154 |
| DCLK or I/O | 171 |
| GND | $7,17,22,32,37,55,65,75,98,103,106,108,118,123,141,152,161$ |
| MODE | 1 |
| PRA or I/O | 148 |
| PRB or I/O | 156 |
| SDI or I/O | 131 |
| $V_{\text {CC }}$ | $12,23,24,27,66,80,107,109,110,113,136,151,166$ |

## Notes:

1. Unused I/O pins aredesignated as outputs by Designer and aredriven LOW.
2. All unassigned pins are available for useas $I / O s$.
3. MODE should beterminated to GND through a 10K resistor to enableActionProbe usage; otherwise, it can be terminated di rectly to GND.

## Package Pin Assignments (continued)

## 196-Pin CQFP (Top View)



| Function | RT1460A, A1460A Pin Number |
| :--- | :--- |
| CLKA or I/O | 172 |
| CLKB or I/O | 173 |
| DCLK or I/O | 196 |
| GND | $1,13,37,51,52,64,86,98,101,112,138,139,149,162,183,193$ |
| HCLK or I/O | 77 |
| IOCLK or I/O | 148 |
| IOPCL or I/O | 100 |
| MODE | 11 |
| PRA or I/O | 174 |
| PRB or I/O | 75 |
| SDI or I/O | 2 |
| V $_{\text {CC }}$ | $12,38,39,59,94,110,111,137,140,155,189$ |

## Notes:

1. Unused I/O pins aredesignated as outputs by ALS and are driven low.
2. All unassigned pins are avai lable for useas $I / O s$.
3. MODE should beterminated to GND through a 10K resistor to enableActi onprobe usage, otherwi seit can be terminated di rectly to GND.

## Package Pin Assignments (continued)

## 256-Pin CQFP (Top View)



| Function | RT14100A, A14100A Pin Number |
| :--- | :--- |
| CLKA or I/O | 219 |
| CLKB or I/O | 220 |
| DCLK or I/O | 256 |
| GND | $1,29,31,59,91,93,110,128,158,160,175,176,189,222,224,240$ |
| HCLK or I/O | 96 |
| IOCLK or I/O | 188 |
| IOPCL or I/O | 127 |
| MODE | 11 |
| PRA or I/O | 225 |
| PRB or I/O | 90 |
| SDI or I/O | 2 |
| V CC | $28,30,46,92,94,141,159,161,174,221,223$ |

## Notes:

1. Unused $\mathrm{I} / \mathrm{O}$ pins aredesignated as outputs by Designer and aredriven LOW.
2. All unassigned pins are available for useas I/Os.
3. MODE should beterminated to GND through a 10 K resistor to enableActionProbe usage; otherwise, it can be terminated di rectly to GND.

## Package Mechanical Drawings

## Ceramic Quad Flatpack (CQFP-Cavity Up)



## Notes:

1. All dimensions arein inches except CQ208 and CQ256 which are in millimeters.
2. Outsideleadframeholes (from dimension H) arecircular for theCQ208 and CQ256.
3. Seal ring and lid are connected to Ground.
4. Lead material is Kovar with minimum 60 mi coniches gold over nickel.
5. Packages are shi pped unformed with the ceramic tie bar.
6. 32200DX - CQ208 has heat sink on the backside.

## Ceramic Quad Flatpack (CQFP)

|  | CQ84 |  |  | CQ132 |  |  | CQ172 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Symbol | Min | Nom. | Max | Min | Nom. | Max | Min | Nom. | Max |
| A | 0.070 | 0.090 | 0.100 | 0.094 | 0.105 | 0.116 | 0.094 | 0.105 | 0.116 |
| A1 | 0.060 | 0.075 | 0.080 | 0.080 | 0.090 | 0.100 | 0.080 | 0.090 | 0.100 |
| b | 0.008 | 0.010 | 0.012 | 0.007 | 0.008 | 0.010 | 0.007 | 0.008 | 0.010 |
| c | 0.004 | 0.006 | 0.008 | 0.004 | 0.006 | 0.008 | 0.004 | 0.006 | 0.008 |
| D1/E1 | 0.640 | 0.650 | 0.660 | 0.940 | 0.950 | 0.960 | 1.168 | 1.180 | 1.192 |
| D2/E2 | 0.500 BSC |  |  | 0.800 BSC |  |  | 1.050 BSC |  |  |
| e | 0.025 BSC |  |  | 0.025 BSC |  |  | 0.025 BSC |  |  |
| F | 0.130 | 0.140 | 0.150 | 0.325 | 0.350 | 0.375 | 0.175 | 0.200 | 0.225 |
| H | 1.460 BSC |  |  | 2.320 BSC |  |  | 2.320 BSC |  |  |
| K | - |  |  | 2.140 BSC |  |  | 2.140 BSC |  |  |
| L1 | 1.595 | 1.600 | 1.615 | 2.485 | 2.500 | 2.505 | 2.485 | 2.495 | 2.505 |

## Note:

1. All dimensions are in inches except CQ256, which is in millimeters.
2. BSC equals Basi c Spacing between Centers. This is a theoretical true position dimension and so has no tolerance.

## Ceramic Quad Flatpack (CQFP)

|  | CQ196 |  |  | CQ256 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Symbol | Min | Nom. | Max | Min | Nom. | Max |
| A | 0.094 | 0.105 | 0.116 | 2.28 | 2.67 | 3.06 |
| A1 | 0.080 | 0.090 | 0.100 | 1.93 | 2.29 | 2.65 |
| b | 0.007 | 0.008 | 0.010 | 0.18 | 0.20 | 0.22 |
| c | 0.004 | 0.006 | 0.008 | 0.11 | 0.15 | 0.18 |
| D1/E1 | 1.336 | 1.350 | 1.364 | 35.64 | 36.00 | 36.36 |
| D2/E2 | 1.200 BSC |  |  | 31.5 BSC |  |  |
| e | 0.025 BSC |  |  | 0.50 BSC |  |  |
| F | 0.175 | 0.200 | 0.225 | 7.05 | 7.75 | 8.45 |
| H | 2.320 BSC |  |  | 70.00 BSC |  |  |
| K | 2.140 BSC |  |  | 65.90 BSC |  |  |
| L1 | 2.485 | 2.495 | 2.505 | 74.60 | 75.00 | 75.40 |

## Note:

1. All dimensions arein inches except CQ256, which is in millimeters.
2. BSC equals Basic Spacing between Centers. This is a theoretical true position dimension and so has no tolerance.

Actel

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http://www.actel.com

Actel Europe Ltd.
Daneshill House, Lutyens Close
Basingstoke, Hampshire RG24 8AG United Kingdom
Tel: +44.(0)1256.305600
Fax: +44.(0)1256.355420

## Actel Corporation

955 East Arques Avenue
Sunnyvale, California 94086 USA
Tel: 408.739.1010
Fax: 408.739.1540

Actel Japan
EXOS Ebisu Bldg. 4F
1-24-14 Ebisu Shibuya-ka
Tokyo 150 Japan
Tel: +81.(0)3.3445.7671
Fax: +81.(0)3.3445.7668


[^0]:    1. Designer Series also supports design entry and simulation tools from Cadence, Mentor Graphics, and Viewlogic.
